



±150°/sec Yaw Rate Gyroscope

ADXRS613

FEATURES

- Complete rate gyroscope on a single chip
- Z-axis (yaw rate) response
- High vibration rejection over wide frequency
- 2000 g powered shock survivability
- Ratiometric to referenced supply
- 5 V single-supply operation
- 105°C operation
- Self-test on digital command
- Ultrasmall and light (<0.15 cc, <0.5 gram)
- Temperature sensor output
- RoHS compliant

APPLICATIONS

- Inertial measurement units
- Platform stabilization
- Robotics

GENERAL DESCRIPTION

The ADXRS613 is a complete angular rate sensor (gyroscope) that uses the Analog Devices, Inc. surface-micromachining process to create a functionally complete and low cost angular rate sensor integrated with all required electronics on one chip. The manufacturing technique for this device is the same high volume BiMOS process used for high reliability automotive airbag accelerometers.

The output signal, RATEOUT (1B, 2A), is a voltage proportional to the angular rate about the axis that is normal to the top surface of the package. The output is ratiometric with respect to a provided reference supply. A single external resistor between SUMJ and RATEOUT can be used to lower the scale factor. An external capacitor sets the bandwidth. Other external capacitors are required for operation.

A temperature output is provided for compensation techniques. Two digital self-test inputs electromechanically excite the sensor to test proper operation of both the sensor and the signal conditioning circuits. The ADXRS613 is available in a 7 mm × 7 mm × 3 mm BGA chip scale package.

FUNCTIONAL BLOCK DIAGRAM

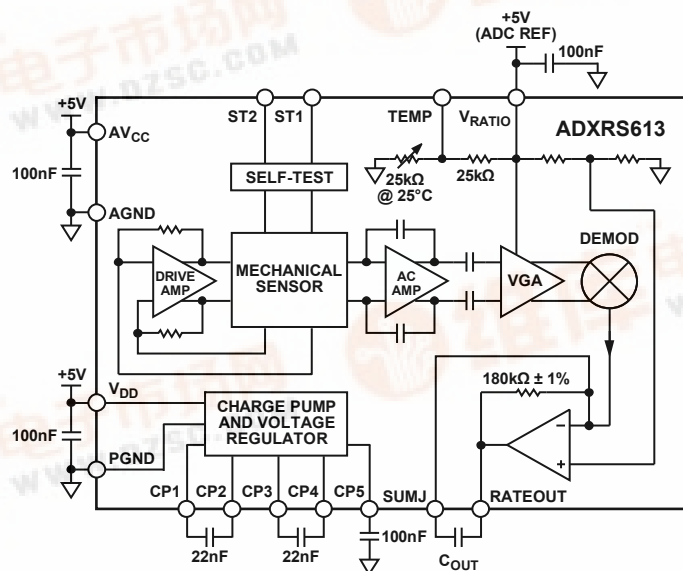


Figure 1.

100-17890



ADXRS613

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REVISION HISTORY

2/08—Revision 0: Initial Version

SPECIFICATIONS

All minimum and maximum specifications are guaranteed. Typical specifications are not guaranteed.

$T_A = -40^{\circ}\text{C}$ to $+105^{\circ}\text{C}$, $V_S = AV_{CC} = V_{DD} = 5\text{ V}$, $V_{\text{RATIO}} = AV_{CC}$, angular rate = $0^{\circ}/\text{sec}$, bandwidth = 80 Hz ($C_{\text{OUT}} = 0.01\text{ }\mu\text{F}$), $I_{\text{OUT}} = 100\text{ }\mu\text{A}$, $\pm 1\text{ g}$, unless otherwise noted.

Table 1.

Parameter	Conditions	ADXRS613BBGZ			Unit
		Min	Typ	Max	
SENSITIVITY (RATIOMETRIC) ¹	Clockwise rotation is positive output				
Measurement Range ²	Full-scale range over specifications range	± 150			$^{\circ}/\text{sec}$
Initial and Over Temperature		11.25	12.5	13.75	$\text{mV}/^{\circ}/\text{sec}$
Temperature Drift ³			± 3		%
Nonlinearity	Best fit straight line		0.1		% of FS
NULL (RATIOMETRIC) ¹					
Null	-40°C to $+105^{\circ}\text{C}$		2.5		V
Null Drift Over Temperature	-40°C to $+105^{\circ}\text{C}$			± 250	mV
Linear Acceleration Effect	Any axis		0.1		$^{\circ}/\text{sec}/\text{g}$
NOISE PERFORMANCE					
Rate Noise Density	$T_A = 25^{\circ}\text{C}$		0.04		$^{\circ}/\text{sec}/\sqrt{\text{Hz}}$
FREQUENCY RESPONSE					
Bandwidth ⁴		1		3000	Hz
Sensor Resonant Frequency			14.5		kHz
SELF-TEST (RATIOMETRIC) ¹					
ST1 RATEOUT Response	ST1 pin from Logic 0 to Logic 1	-500	-1000		mV
ST2 RATEOUT Response	ST2 pin from Logic 0 to Logic 1	500	1000		mV
Logic 1 Input Voltage		$0.8 \times V_{\text{RATIO}}$			V
Logic 0 Input Voltage				$0.2 \times V_{\text{RATIO}}$	V
Input Impedance	To common		50		k Ω
TEMPERATURE SENSOR (RATIOMETRIC) ¹					
V_{OUT} at 25°C	Load = 100 M Ω	2.35	2.5	2.65	V
Scale Factor ⁵	@ 25°C , $V_{\text{RATIO}} = 5\text{ V}$		9.1		$\text{mV}/^{\circ}\text{C}$
Load to V_S			25		k Ω
Load to Common			25		k Ω
TURN-ON TIME	Power on to $\pm 1/2^{\circ}/\text{sec}$ of final			50	ms
OUTPUT DRIVE CAPABILITY					
Current Drive	For rated specifications			200	μA
Capacitive Load Drive				1000	pF
POWER SUPPLY					
Operating Voltage (V_S)		4.75	5.00	5.25	V
V_{RATIO} Input		3		V_S	V
Supply Current			3.5	5.0	mA
TEMPERATURE RANGE					
Specified Performance		-40		$+105$	$^{\circ}\text{C}$

¹ Parameter is linearly ratiometric with V_{RATIO} .

² The maximum range possible, including output swing range, initial offset, sensitivity, offset drift, and sensitivity drift at 5 V supplies.

³ From $+25^{\circ}\text{C}$ to -40°C or from $+25^{\circ}\text{C}$ to $+105^{\circ}\text{C}$.

⁴ Adjusted by external capacitor, C_{OUT} .

⁵ For a change in temperature from 25°C to 26°C . V_{TEMP} is ratiometric to V_{RATIO} . See the Temperature Output and Calibration section for more details.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

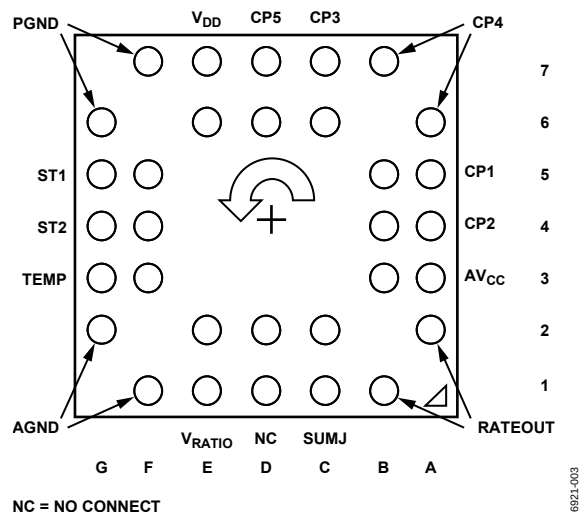


Figure 3. Pin Configuration

Table 3. Pin Function Descriptions

Pin No.	Mnemonic	Description
6D, 7D	CP5	HV Filter Capacitor (100 nF).
6A, 7B	CP4	Charge Pump Capacitor (22 nF).
6C, 7C	CP3	Charge Pump Capacitor (22 nF).
5A, 5B	CP1	Charge Pump Capacitor (22 nF).
4A, 4B	CP2	Charge Pump Capacitor (22 nF).
3A, 3B	AV _{CC}	Positive Analog Supply.
1B, 2A	RATEOUT	Rate Signal Output.
1C, 2C	SUMJ	Output Amplifier Summing Junction.
1D, 2D	NC	No Connect.
1E, 2E	V _{RATIO}	Reference Supply for Ratiometric Output.
1F, 2G	AGND	Analog Supply Return.
3F, 3G	TEMP	Temperature Voltage Output.
4F, 4G	ST2	Self-Test for Sensor 2.
5F, 5G	ST1	Self-Test for Sensor 1.
6G, 7F	PGND	Charge Pump Supply Return.
6E, 7E	V _{DD}	Positive Charge Pump Supply.

TYPICAL PERFORMANCE CHARACTERISTICS

N > 1000 for all typical performance plots, unless otherwise noted.

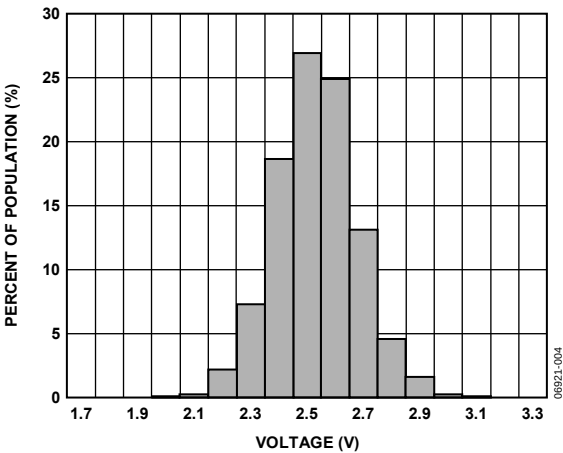


Figure 4. Null Output at 25°C ($V_{RATIO} = 5\text{ V}$)

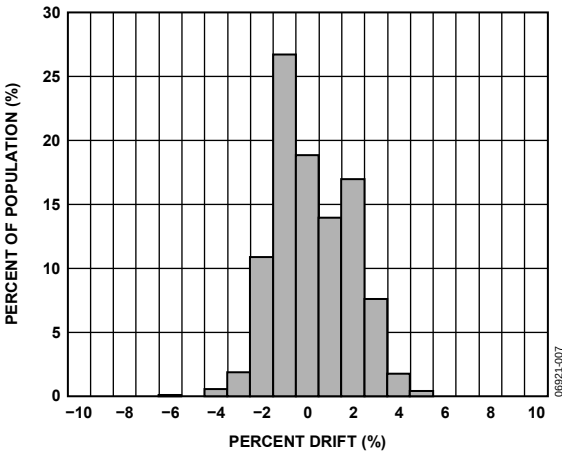


Figure 7. Sensitivity Drift over Temperature

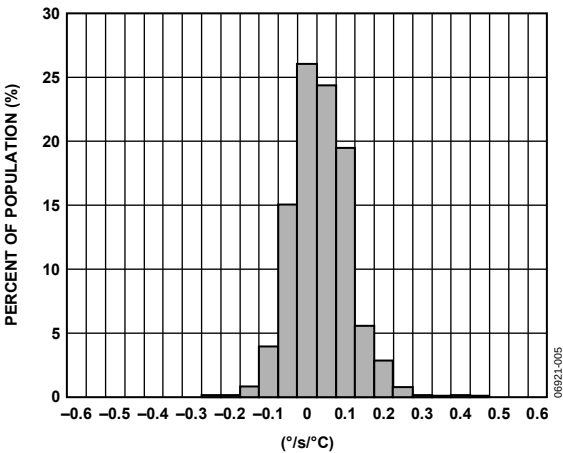


Figure 5. Null Drift over Temperature ($V_{RATIO} = 5\text{ V}$)

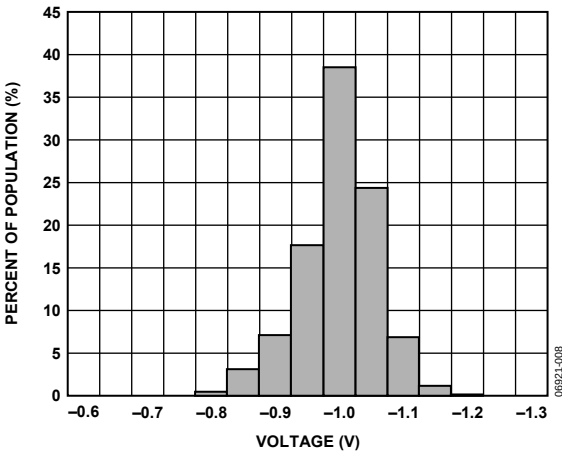


Figure 8. ST1 Output Change at 25°C ($V_{RATIO} = 5\text{ V}$)

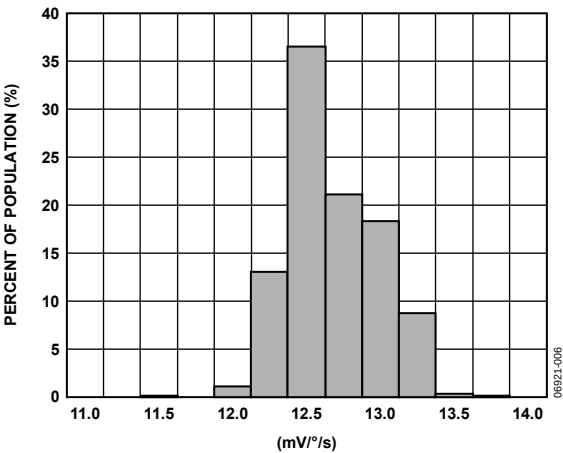


Figure 6. Sensitivity at 25°C ($V_{RATIO} = 5\text{ V}$)

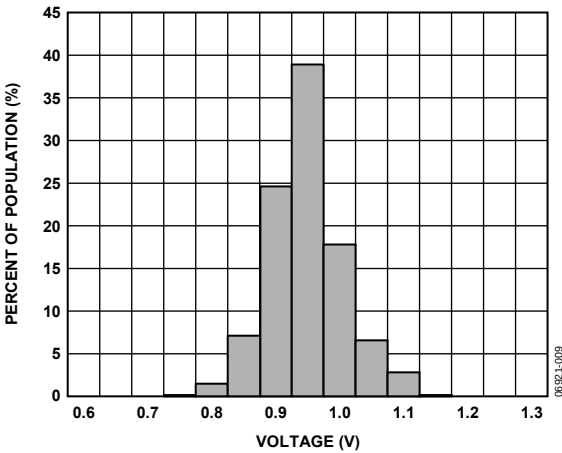


Figure 9. ST2 Output Change at 25°C ($V_{RATIO} = 5\text{ V}$)

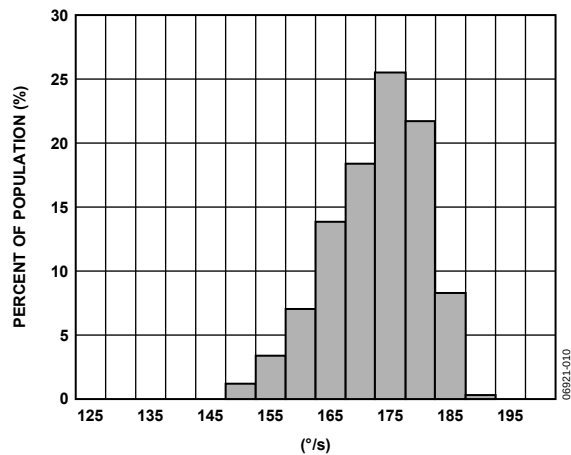


Figure 10. Measurement Range

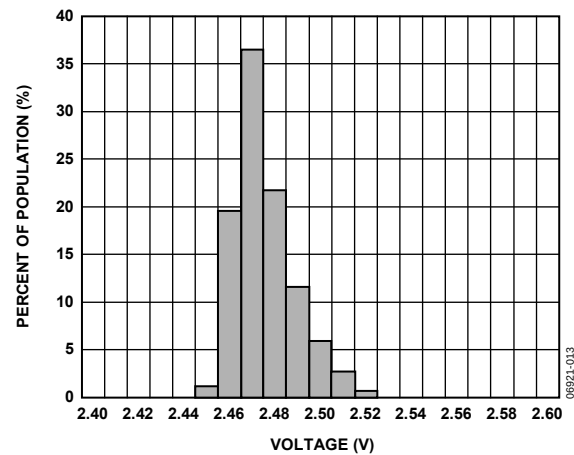


Figure 13. V_{TEMP} Output at 25°C ($V_{RATIO} = 5$ V)

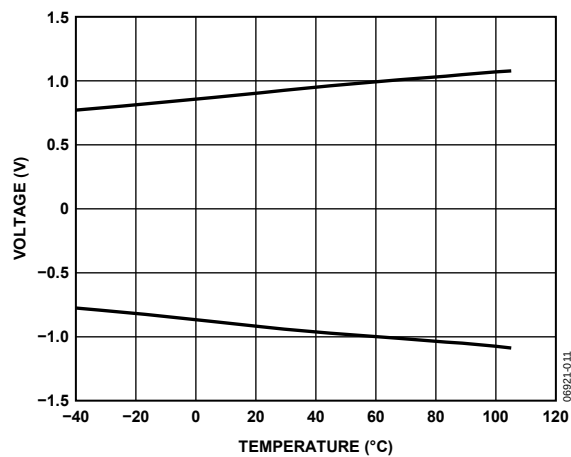


Figure 11. Typical Self-Test Change over Temperature

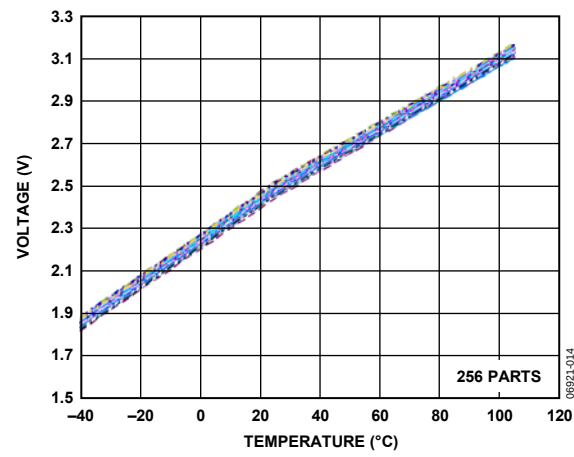


Figure 14. V_{TEMP} Output over Temperature ($V_{RATIO} = 5$ V)

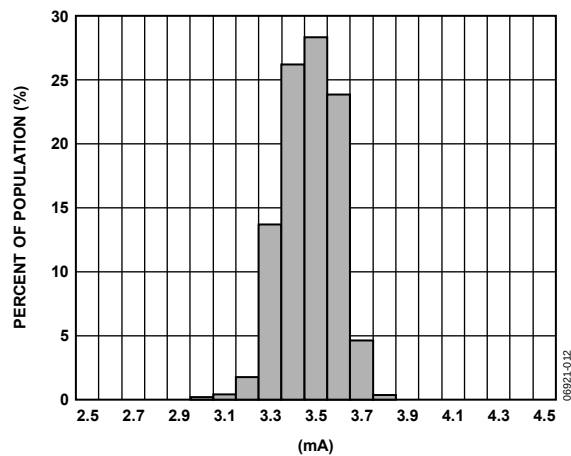


Figure 12. Current Consumption at 25°C ($V_{RATIO} = 5$ V)

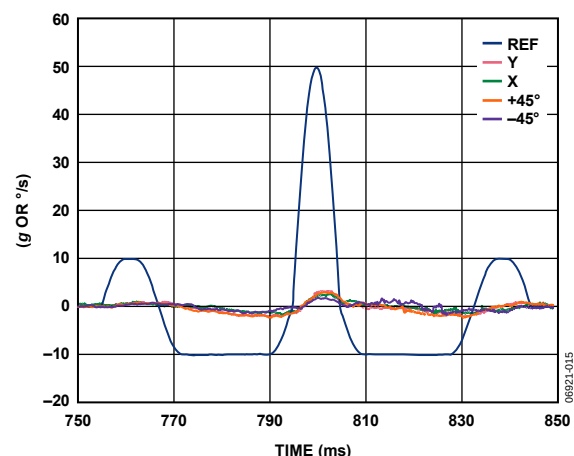


Figure 15. g and $g \times g$ Sensitivity for a 50 g, 10 ms Pulse

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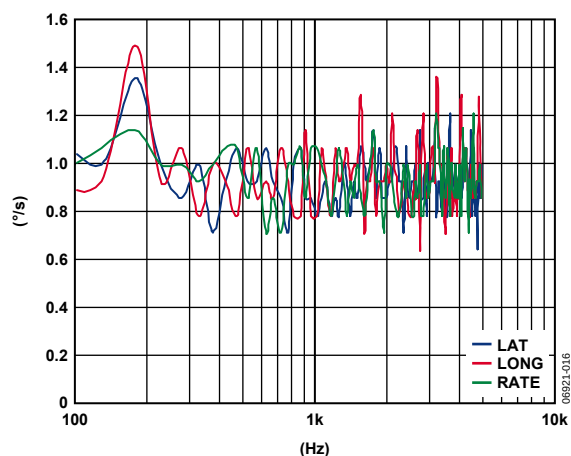


Figure 16. Typical Response to 10 g Sinusoidal Vibration (Sensor Bandwidth = 2 kHz)

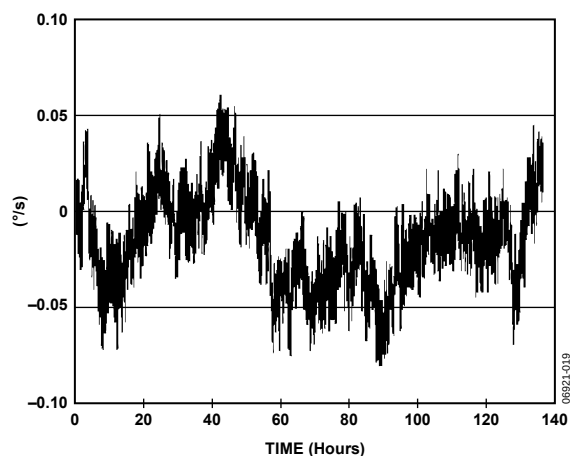


Figure 19. Typical Shift in 90 sec Null Averages Accumulated over 140 Hours

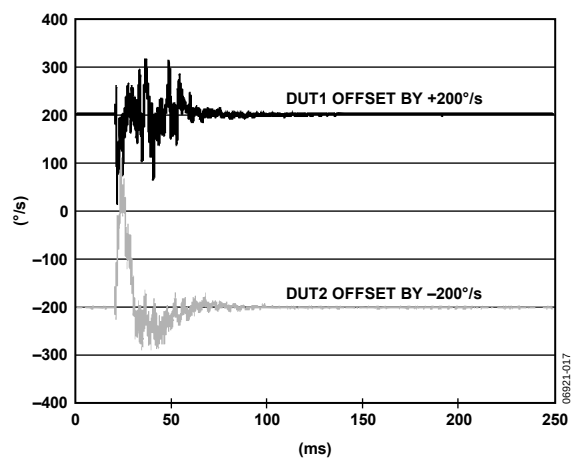


Figure 17. Typical High g (2500 g) Shock Response (Sensor Bandwidth = 40 Hz)

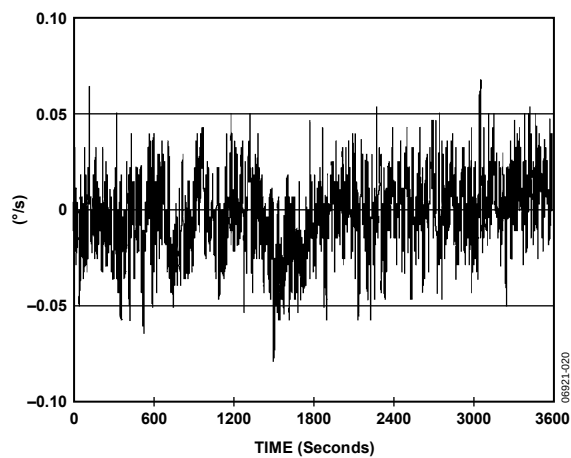


Figure 20. Typical Shift in Short-Term Null (Bandwidth = 1 Hz)

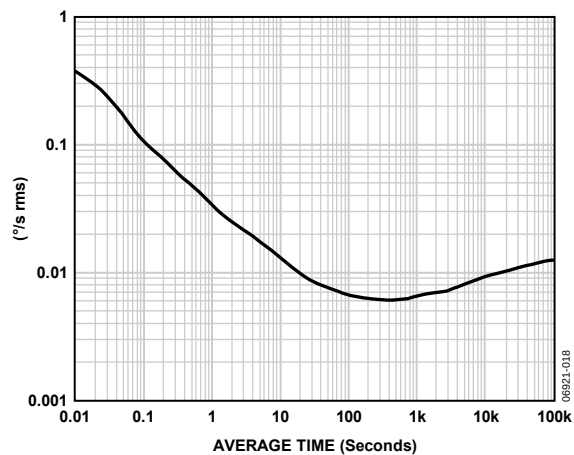


Figure 18. Typical Root Allan Deviation at 25°C vs. Averaging Time

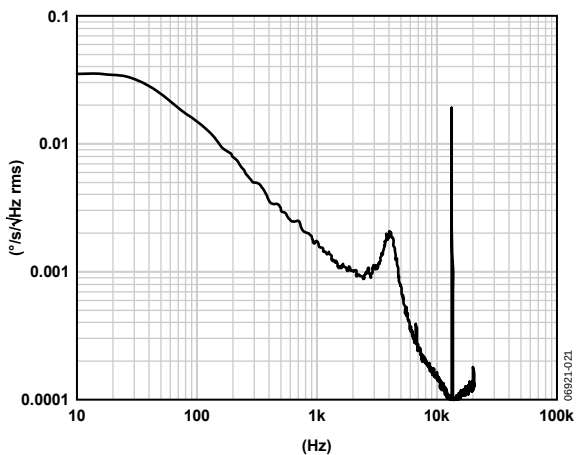


Figure 21. Typical Noise Spectral Density (Bandwidth = 40 Hz)

THEORY OF OPERATION

The ADXRS613 operates on the principle of a resonator gyroscope. Two polysilicon sensing structures each contain a dither frame that is electrostatically driven to resonance, producing the necessary velocity element to produce a Coriolis force while rotating. At two of the outer extremes of each frame, orthogonal to the dither motion, are movable fingers that are placed between fixed pickoff fingers to form a capacitive pickoff structure that senses Coriolis motion. The resulting signal is fed to a series of gain and demodulation stages that produce the electrical rate signal output. The dual-sensor design rejects external g forces and vibration. Fabricating the sensor with the signal conditioning electronics preserves signal integrity in noisy environments.

The electrostatic resonator requires 18 V to 20 V for operation. Because only 5 V are typically available in most applications, a charge pump is included on-chip. If an external 18 V to 20 V supply is available, the two capacitors on CP1 through CP4 can be omitted, and this supply can be connected to the CP5 pin (6D, 7D). Note that CP5 should not be grounded when power is applied to the ADXRS613. Although no damage occurs, under certain conditions, the charge pump may fail to start up after the ground is removed without first removing power from the ADXRS613.

SETTING BANDWIDTH

External Capacitor C_{OUT} is used in combination with the on-chip R_{OUT} resistor to create a low-pass filter to limit the bandwidth of the ADXRS613 rate response. The -3 dB frequency set by R_{OUT} and C_{OUT} is

$$f_{OUT} = \frac{1}{(2 \times \pi \times R_{OUT} \times C_{OUT})}$$

and can be well controlled because R_{OUT} has been trimmed during manufacturing to be $180 \text{ k}\Omega \pm 1\%$. Any external resistor applied between the RATEOUT pin (1B, 2A) and SUMJ pin (1C, 2C) results in:

$$R_{OUT} = \frac{(180 \text{ k}\Omega \times R_{EXT})}{(180 \text{ k}\Omega + R_{EXT})}$$

In general, an additional hardware or software filter is added to attenuate high frequency noise arising from demodulation spikes at the gyroscope's 14 kHz resonant frequency (the noise spikes at 14 kHz can be clearly seen in the power spectral density curve shown in Figure 21). Typically, this additional filter's corner frequency is set to greater than 5× the required bandwidth to preserve good phase response.

Figure 22 shows the effect of adding a 250 Hz filter to the output of an ADXRS613 set to 40 Hz bandwidth (as shown in Figure 21). High frequency demodulation artifacts are attenuated by approximately 18 dB.

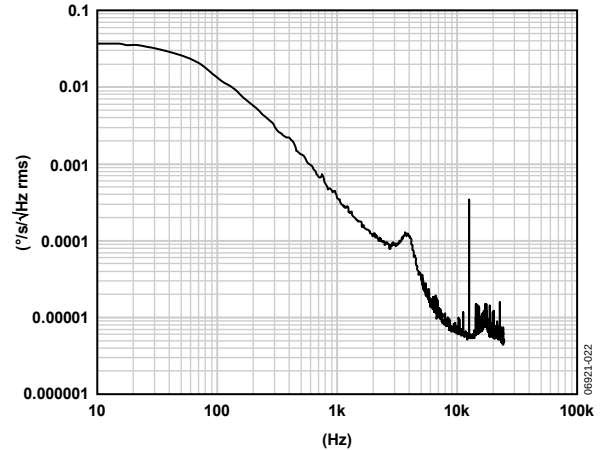


Figure 22. Noise Spectral Density with Additional 250 Hz Filter

TEMPERATURE OUTPUT AND CALIBRATION

It is common practice to temperature-calibrate gyroscopes to improve their overall accuracy. The ADXRS613 has a temperature proportional voltage output that provides input to such a calibration method. The temperature sensor structure is shown in Figure 23. The temperature output is characteristically nonlinear, and any load resistance connected to the TEMP output results in decreasing the TEMP output and temperature coefficient. Therefore, buffering the output is recommended.

The voltage at the TEMP pin (3F, 3G) is nominally 2.5 V at 25°C and $V_{RATIO} = 5 \text{ V}$. The temperature coefficient is $\sim 9 \text{ mV}/^\circ\text{C}$ at 25°C. Although the TEMP output is highly repeatable, it has only modest absolute accuracy.

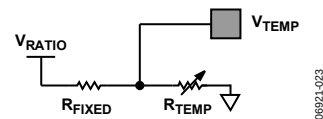


Figure 23. ADXRS613 Temperature Sensor Structure

CALIBRATED PERFORMANCE

Using a 3-point calibration technique, it is possible to calibrate the null and sensitivity drift of the ADXRS613 to an overall accuracy of nearly 200°/hour. An overall accuracy of 40°/hour or better is possible using more points.

Limiting the bandwidth of the device reduces the flat-band noise during the calibration process, improving the measurement accuracy at each calibration point.

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ADXRS613 AND SUPPLY RATIOMETRICITY

The ADXRS613 RATEOUT and TEMP signals are ratiometric to the V_{RATIO} voltage, that is, the null voltage, rate sensitivity, and temperature outputs are proportional to V_{RATIO} . Thus, the ADXRS613 is most easily used with a supply-ratiometric ADC that results in self-cancellation of errors due to minor supply variations. There is some small error due to nonratiometric behavior. Typical ratiometricity error for null, sensitivity, self-test, and temperature output is outlined in Table 4.

Note that V_{RATIO} must never be greater than AV_{CC} .

Table 4. Ratiometricity Error for Various Parameters

Parameter	$V_S = V_{\text{RATIO}} = 4.75 \text{ V}$	$V_S = V_{\text{RATIO}} = 5.25 \text{ V}$
ST1		
Mean	−0.4%	−0.3%
Sigma	0.6%	0.6%
ST2		
Mean	−0.4%	−0.3%
Sigma	0.6%	0.6%
Null		
Mean	−0.04%	−0.02%
Sigma	0.3%	0.2%
Sensitivity		
Mean	0.03%	0.1%
Sigma	0.1%	0.1%
V_{TEMP}		
Mean	−0.3%	−0.5%
Sigma	0.1%	0.1%

NULL ADJUSTMENT

The nominal 2.5 V null is for a symmetrical swing range at RATEOUT (1B, 2A). However, a nonsymmetrical output swing may be suitable in some applications. Null adjustment is possible by injecting a suitable current to SUMJ (1C, 2C). Note that supply disturbances may reflect some null instability. Digital supply noise should be avoided particularly in this case.

SELF-TEST FUNCTION

The ADXRS613 includes a self-test feature that actuates each of the sensing structures and associated electronics as if subjected to angular rate. It is activated by standard logic high levels applied to Input ST1 (5F, 5G), Input ST2 (4F, 4G), or both. ST1 causes the voltage at RATEOUT to change about −1.9 V, and ST2 causes an opposite change of +1.9 V. The self-test response follows the viscosity temperature dependence of the package atmosphere, approximately 0.25%/°C.

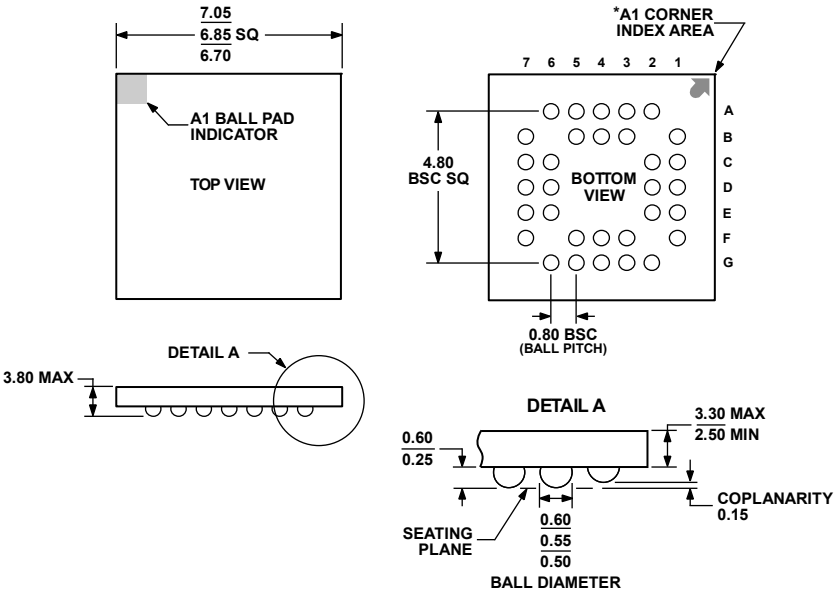
Activating both ST1 and ST2 simultaneously is not damaging. ST1 and ST2 are fairly closely matched ($\pm 5\%$), but actuating both simultaneously may result in a small apparent null bias shift proportional to the degree of self-test mismatch.

ST1 and ST2 are activated by applying a voltage of greater than $0.8 \times V_{\text{RATIO}}$ to the ST1 and ST2 pins. ST1 and ST2 are deactivated by applying a voltage of less than $0.2 \times V_{\text{RATIO}}$ to the ST1 and ST2 pins. The voltage applied to ST1 and ST2 must never be greater than AV_{CC} .

CONTINUOUS SELF-TEST

The one-chip integration of the ADXRS613 gives it higher reliability than is obtainable with any other high volume manufacturing method. In addition, it is manufactured under a mature BiMOS process with field-proven reliability. As an additional failure detection measure, a power-on self-test can be performed. However, some applications may warrant continuous self-test while sensing rate. Details outlining continuous self-test techniques are also available in the AN-768 Application Note, *Using the ADXRS150/ADXRS300 in Continuous Self-Test Mode* at www.analog.com.

OUTLINE DIMENSIONS



*BALL A1 IDENTIFIER IS GOLD PLATED AND CONNECTED TO THE D/A PAD INTERNALLY VIA HOLES.

Figure 24. 32-Lead Ceramic Ball Grid Array [CBGA] (BG-32-3)

Dimensions shown in millimeters

060506-A

ORDERING GUIDE

Model	Temperature Range	Package Description	Package Option
ADXRS613BBGZ ¹	–40°C to +105°C	32-Lead Ceramic Ball Grid Array (CBGA)	BG-32-3
ADXRS613BBGZ-RL ¹	–40°C to +105°C	32-Lead Ceramic Ball Grid Array (CBGA)	BG-32-3
EVAL-ADXRS613Z ¹		Evaluation Board	

¹ Z = RoHS Compliant Part.

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NOTES