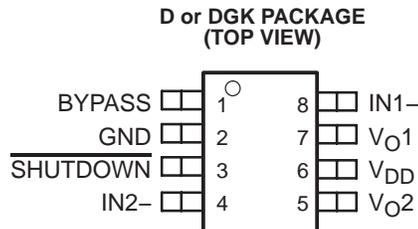


TPA6102A2

50-mW ULTRALOW-VOLTAGE, FIXED-GAIN STEREO HEADPHONE AUDIO POWER AMPLIFIER

SLOS324B – JUNE 2000 – REVISED SEPTEMBER 2004

- 50-mW Stereo Output
- Low Supply Current . . . 0.75 mA
- Low Shutdown Current . . . 50 nA
- Minimal External Components Required
- Gain Set Internally to 14 dB
- Pop Reduction Circuitry
- Internal Mid-Rail Generation
- Thermal and Short-Circuit Protection
- Surface-Mount Packaging
 - MSOP
 - SOIC
- 1.6-V to 3.6-V Supply Voltage Range



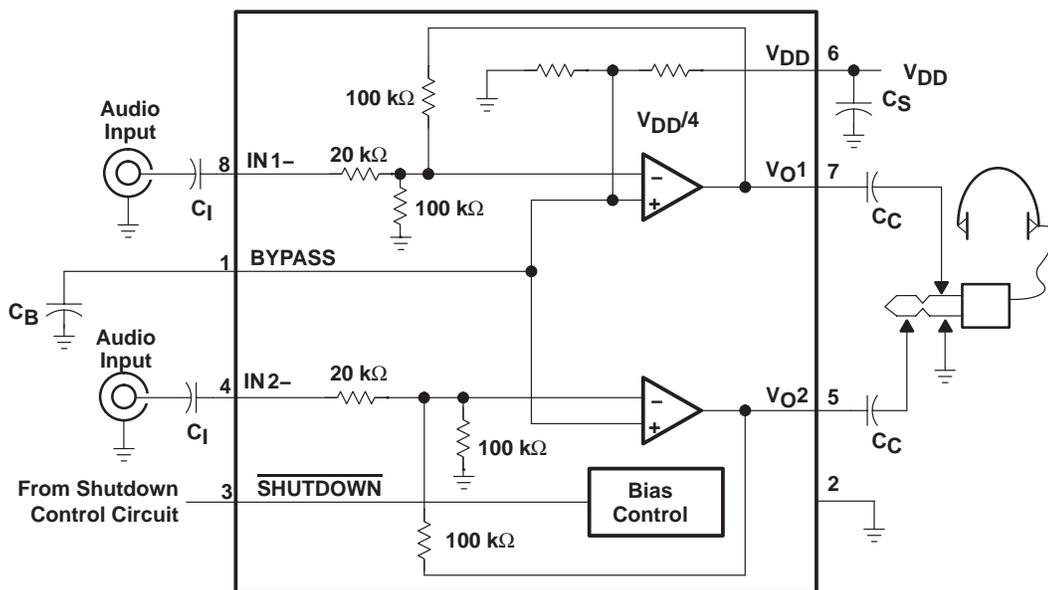
description

The TPA6102A2 is a stereo audio power amplifier packaged in either an 8-pin SOIC package or an 8-pin MOSP package capable of delivering 50 mW of continuous RMS power per channel into 16-Ω loads. Amplifier gain is internally set to 14 dB (inverting) to save board space by eliminating six external resistors.

The TPA6102A2 is optimized for battery applications because of its low-supply current, shutdown current, and THD+N. To obtain the low-supply voltage range, the TPA6102A2 biases BYPASS to $V_{DD}/4$.

When driving a 16-Ω load with 40-mW output power from 3.3 V, THD+N is 0.08% at 1 kHz, and less than 0.2% across the audio band of 20 Hz to 20 kHz. For 30 mW into 32-Ω loads, the THD+N is reduced to less than 0.06% at 1 kHz, and is less than 0.3% across the audio band of 20 Hz to 20 kHz.

typical application circuit



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

**TEXAS
INSTRUMENTS**

POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

Copyright © 2004, Texas Instruments Incorporated

TPA6102A2
50-mW ULTRALOW-VOLTAGE, FIXED-GAIN STEREO HEADPHONE
AUDIO POWER AMPLIFIER

SLOS324B – JUNE 2000 – REVISED SEPTEMBER 2004

AVAILABLE OPTIONS

T _A	PACKAGED DEVICE		MSOP SYMBOLIZATION
	SMALL OUTLINE (D)	MSOP (DGK)	
-40°C to 85°C	TPA6102A2D	TPA6102A2DGK	AJN

Terminal Functions

TERMINAL NAME	NO.	I/O	DESCRIPTION
BYPASS	1	I	Tap to voltage divider for internal mid-supply bias supply. BYPASS is set at V _{DD} /4. Connect to a 0.1-μF to 1-μF low ESR capacitor for best performance.
GND	2	I	GND is the ground connection.
IN1-	8	I	IN1- is the inverting input for channel 1.
IN2-	4	I	IN2- is the inverting input for channel 2.
SHUTDOWN	3	I	Active-low input. When held low, the device is placed in a low supply current mode.
V _{DD}	6	I	V _{DD} is the supply voltage terminal.
V _{O1}	7	O	V _{O1} is the audio output for channel 1.
V _{O2}	5	O	V _{O2} is the audio output for channel 2.

absolute maximum ratings over operating free-air temperature (unless otherwise noted)†

Supply voltage, V _{DD}	4 V
Input voltage, V _I	-0.3 V to V _{DD} + 0.3 V
Continuous total power dissipation	Internally Limited
Operating junction temperature range, T _J	-40°C to 150°C
Storage temperature range, T _{stg}	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

DISSIPATION RATING TABLE

PACKAGE	T _A ≤ 25°C POWER RATING	DERATING FACTOR ABOVE T _A = 25°C	T _A = 70°C POWER RATING	T _A = 85°C POWER RATING
D	710 mW	5.68 mW/°C	454 mW	369 mW
DGK	469 mW	3.75 mW/°C	300 mW	244 mW

recommended operating conditions

	MIN	MAX	UNIT
Supply voltage, V _{DD}	1.6	3.6	V
High-level input voltage, V _{IH} (SHUTDOWN)	60% x V _{DD}		V
Low-level input voltage, V _{IL} (SHUTDOWN)	25% x V _{DD}		V
Operating free-air temperature, T _A	-40	85	°C



TPA6102A2
50-mW ULTRALOW-VOLTAGE, FIXED-GAIN STEREO HEADPHONE
AUDIO POWER AMPLIFIER

SLOS324B – JUNE 2000 – REVISED SEPTEMBER 2004

dc electrical characteristics at $T_A = 25^\circ\text{C}$, $V_{DD} = 3.6\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{OO}	Output offset voltage	$A_V = 14\text{ dB}$		5	40	mV
PSRR	Power supply rejection ratio	$V_{DD} = 3\text{ V to }3.6\text{ V}$		72		dB
I_{DD}	Supply current	$\overline{\text{SHUTDOWN}} = 3.6\text{ V}$		0.75	1.5	mA
$I_{DD}(\text{SD})$	Supply current in $\overline{\text{SHUTDOWN}}$ mode	$\overline{\text{SHUTDOWN}} = 0\text{ V}$		50	250	nA
$ I_{IH} $	High-level input current ($\overline{\text{SHUTDOWN}}$)	$V_{DD} = 3.6\text{ V}$, $V_I = V_{DD}$			1	μA
$ I_{IL} $	Low-level input current ($\overline{\text{SHUTDOWN}}$)	$V_{DD} = 3.6\text{ V}$, $V_I = 0\text{ V}$			1	μA
Z_I	Input impedance			20		$\text{k}\Omega$

ac operating characteristics, $V_{DD} = 3.3\text{ V}$, $T_A = 25^\circ\text{C}$, $R_L = 16\ \Omega$

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
G	Gain			14		dB
P_O	Output power (each channel)	$\text{THD} \leq 0.1\%$, $f = 1\text{ kHz}$		50		mW
THD+N	Total harmonic distortion + noise	$P_O = 45\text{ mW}$, $20\text{--}20\text{ kHz}$		0.4%		
B_{OM}	Maximum output power BW	$\text{THD} < 0.5\%$		> 20		kHz
k_{SVR}	Supply ripple rejection ratio	$f = 1\text{ kHz}$		47		dB
SNR	Signal-to-noise ratio	$P_O = 50\text{ mW}$		86		dB
V_n	Noise output voltage (no noise weighting filter)			45		$\mu\text{V}(\text{rms})$

ac operating characteristics, $V_{DD} = 3.3\text{ V}$, $T_A = 25^\circ\text{C}$, $R_L = 32\ \Omega$

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
G	Gain			14		dB
P_O	Output power (each channel)	$\text{THD} \leq 0.1\%$, $f = 1\text{ kHz}$		35		mW
THD+N	Total harmonic distortion + noise	$P_O = 30\text{ mW}$, $20\text{--}20\text{ kHz}$		0.4%		
B_{OM}	Maximum output power BW	$\text{THD} < 0.4\%$		> 20		kHz
k_{SVR}	Supply ripple rejection ratio	$f = 1\text{ kHz}$		47		dB
SNR	Signal-to-noise ratio	$P_O = 30\text{ mW}$		86		dB
V_n	Noise output voltage (no noise weighting filter)			50		$\mu\text{V}(\text{rms})$

TPA6102A2

50-mW ULTRALOW-VOLTAGE, FIXED-GAIN STEREO HEADPHONE AUDIO POWER AMPLIFIER

SLOS324B – JUNE 2000 – REVISED SEPTEMBER 2004

dc electrical characteristics at $T_A = 25^\circ\text{C}$, $V_{DD} = 1.6\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{OO}	Output offset voltage	$A_V = 14\text{ dB}$		5	40	mV
PSRR	Power supply rejection ratio	$V_{DD} = 1.4\text{ V to } 1.8\text{ V}$		80		dB
I_{DD}	Supply current	$\overline{\text{SHUTDOWN}} = 1.6\text{ V}$		0.65	1.2	mA
$I_{DD}(\text{SD})$	Supply current in SHUTDOWN mode	$\overline{\text{SHUTDOWN}} = 0\text{ V}$		50	250	nA
$ I_{IH} $	High-level input current ($\overline{\text{SHUTDOWN}}$)	$V_{DD} = 1.6\text{ V}, V_I = V_{DD}$			1	μA
$ I_{IL} $	Low-level input current ($\overline{\text{SHUTDOWN}}$)	$V_{DD} = 1.6\text{ V}, V_I = 0\text{ V}$			1	μA
Z_I	Input impedance			20		k Ω

ac operating characteristics, $V_{DD} = 1.6\text{ V}$, $T_A = 25^\circ\text{C}$, $R_L = 16\ \Omega$

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
G	Gain			14		dB
P_O	Output power (each channel)	THD $\leq 0.5\%$, $f = 1\text{ kHz}$		10		mW
THD+N	Total harmonic distortion + noise	$P_O = 9.5\text{ mW}$, 20–20 kHz		0.06%		
B_{OM}	Maximum output power BW	THD $< 1\%$		> 20		kHz
k_{SVR}	Supply ripple rejection ratio	$f = 1\text{ kHz}$		47		dB
SNR	Signal-to-noise ratio	$P_O = 10\text{ mW}$		82		dB
V_n	Noise output voltage (no noise weighting filter)			32		$\mu\text{V}(\text{rms})$

ac operating characteristics, $V_{DD} = 1.6\text{ V}$, $T_A = 25^\circ\text{C}$, $R_L = 32\ \Omega$

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
G	Gain			14		dB
P_O	Output power (each channel)	THD $\leq 0.5\%$, $f = 1\text{ kHz}$		7.5		mW
THD+N	Total harmonic distortion + noise	$P_O = 6.5\text{ mW}$, 20–20 kHz		0.05%		
B_{OM}	Maximum output power BW	THD $< 1\%$		> 20		kHz
k_{SVR}	Supply ripple rejection ratio	$f = 1\text{ kHz}$		47		dB
SNR	Signal-to-noise ratio	$P_O = 7.5\text{ mW}$		84		dB
V_n	Noise output voltage (no noise weighting filter)			32		$\mu\text{V}(\text{rms})$

TYPICAL CHARACTERISTICS

Table of Graphs

		FIGURE
THD+N	Total harmonic distortion plus noise	vs Frequency
		vs Output power
		vs Output voltage
P_O	Output power	vs Load resistance
k_{SVR}	Supply ripple rejection ratio	vs Frequency
V_n	Output noise voltage	vs Frequency
	Crosstalk	vs Frequency
	Closed-loop gain and phase	vs Frequency
I_{DD}	Supply current	vs Supply voltage
P_D	Power dissipation	vs Output power



TPA6102A2
50-mW ULTRALOW-VOLTAGE, FIXED-GAIN STEREO HEADPHONE
AUDIO POWER AMPLIFIER

SLOS324B – JUNE 2000 – REVISED SEPTEMBER 2004

TYPICAL CHARACTERISTICS

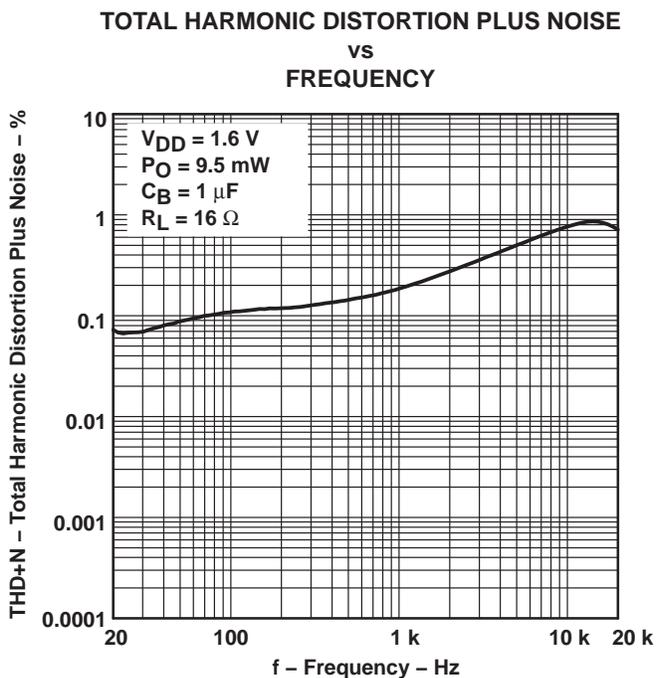


Figure 1

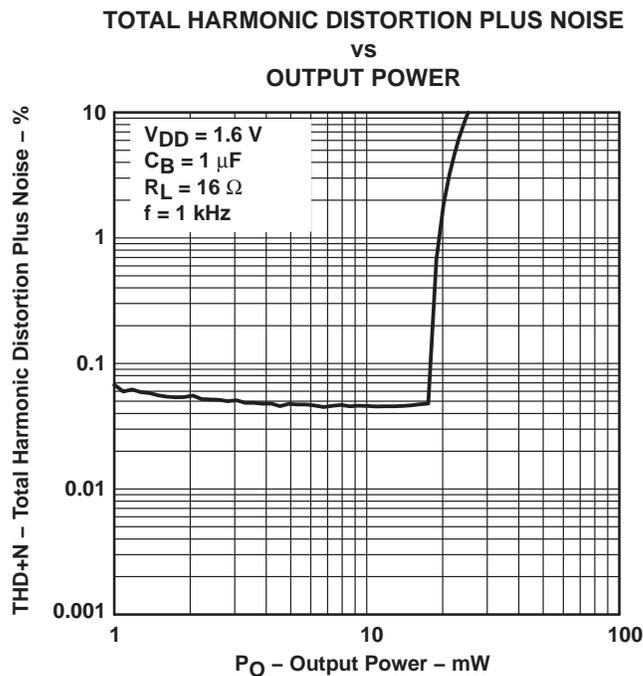


Figure 2

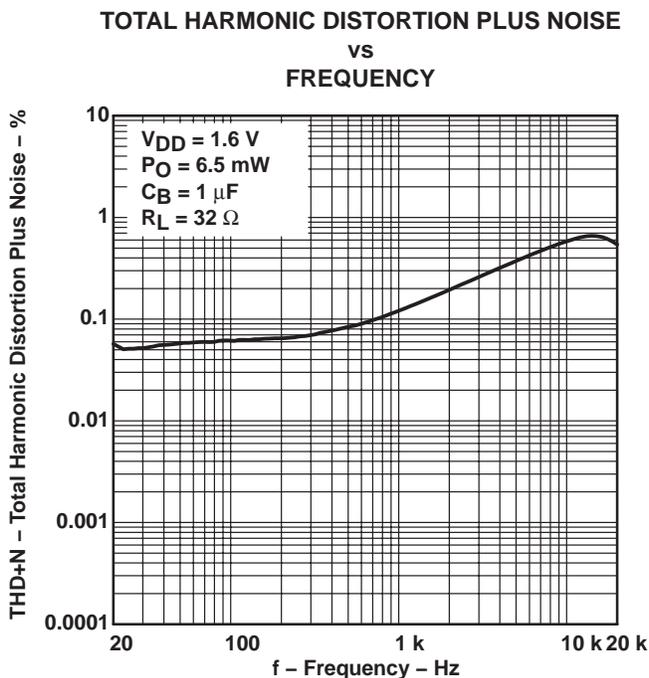


Figure 3

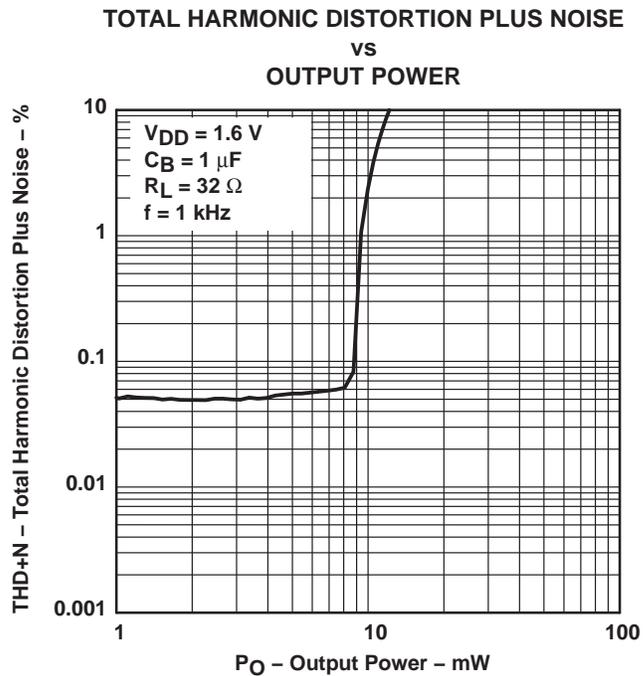


Figure 4

TYPICAL CHARACTERISTICS

TOTAL HARMONIC DISTORTION PLUS NOISE
VS
FREQUENCY

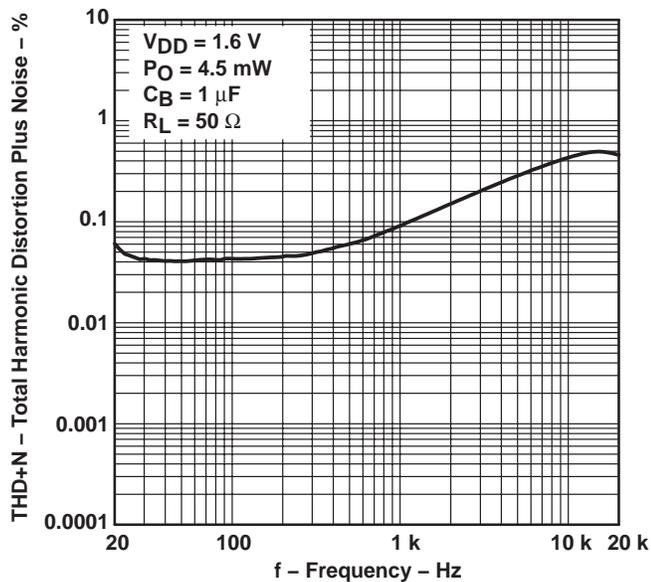


Figure 5

TOTAL HARMONIC DISTORTION PLUS NOISE
VS
OUTPUT POWER

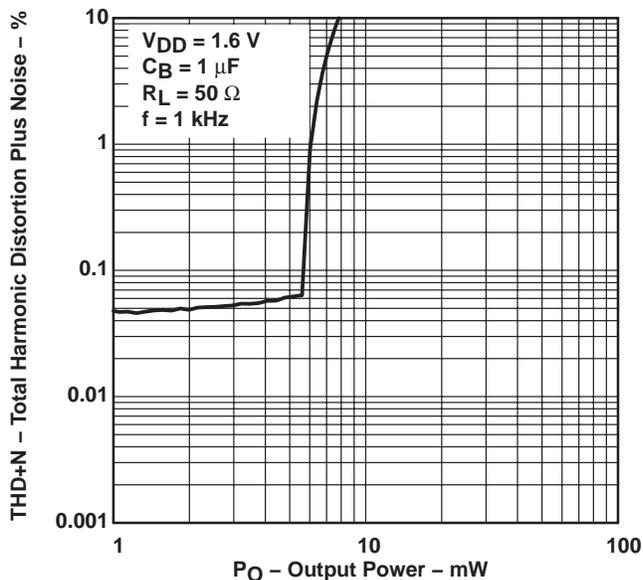


Figure 6

TOTAL HARMONIC DISTORTION PLUS NOISE
VS
FREQUENCY

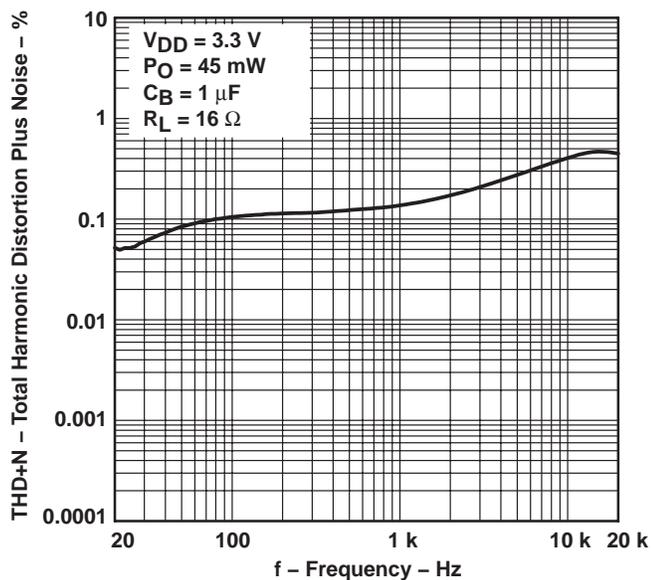


Figure 7

TOTAL HARMONIC DISTORTION PLUS NOISE
VS
OUTPUT POWER

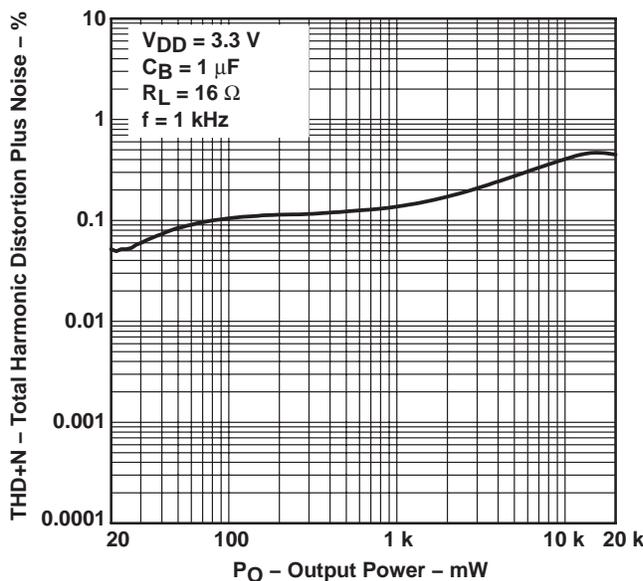


Figure 8

TPA6102A2
50-mW ULTRALOW-VOLTAGE, FIXED-GAIN STEREO HEADPHONE
AUDIO POWER AMPLIFIER

SLOS324B – JUNE 2000 – REVISED SEPTEMBER 2004

TYPICAL CHARACTERISTICS

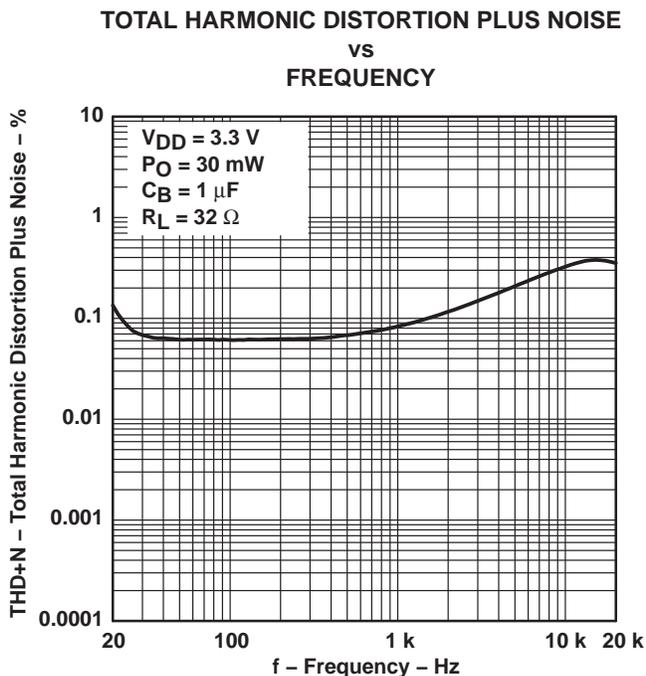


Figure 9

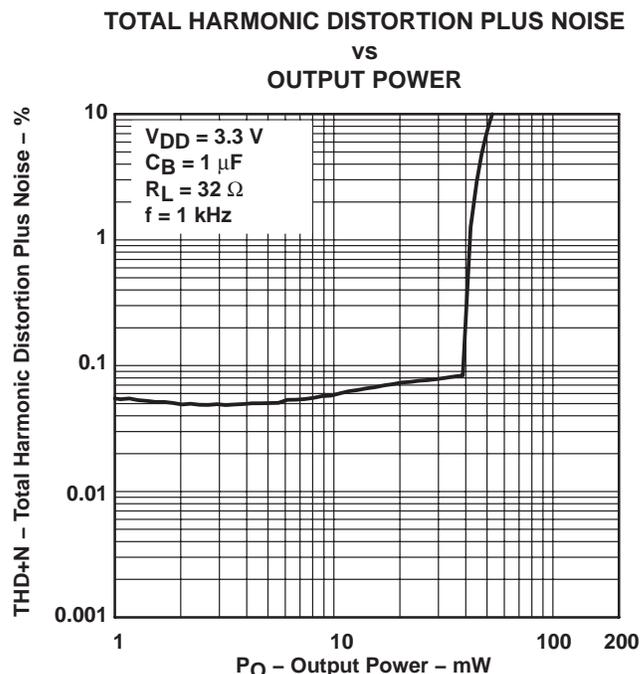


Figure 10

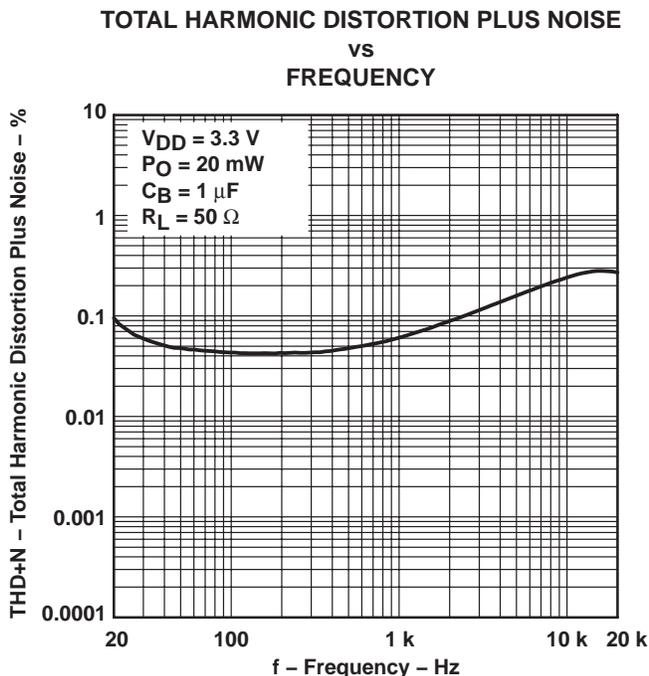


Figure 11

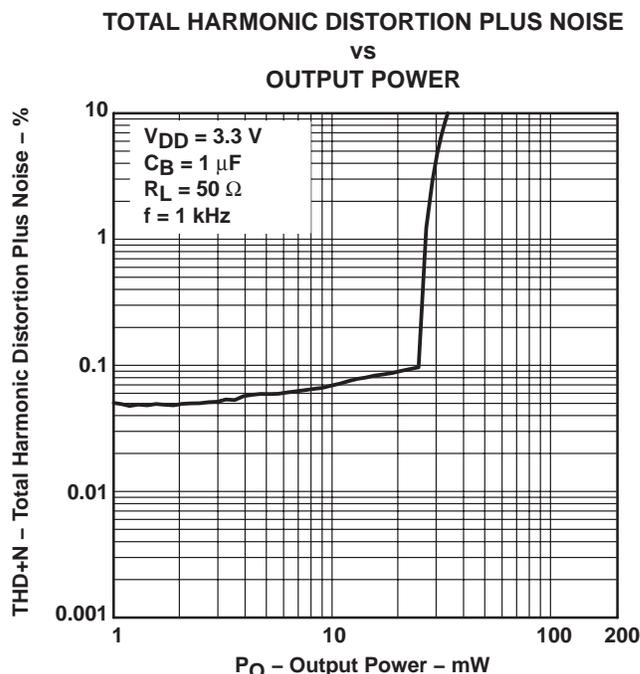


Figure 12

TYPICAL CHARACTERISTICS

TOTAL HARMONIC DISTORTION PLUS NOISE
vs
OUTPUT VOLTAGE

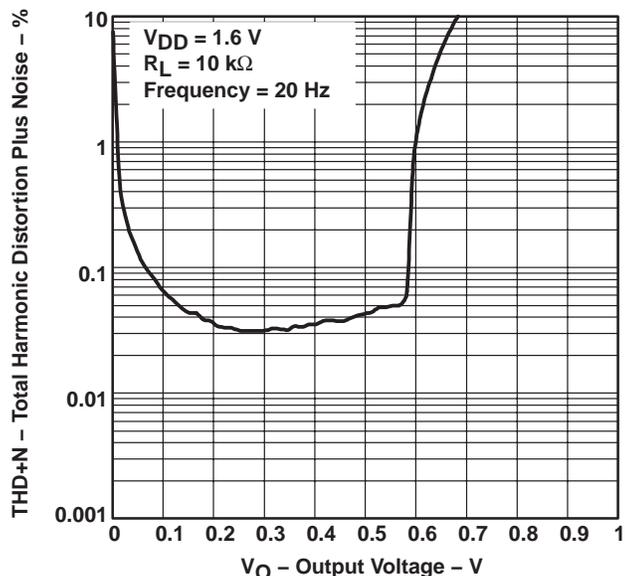


Figure 13

TOTAL HARMONIC DISTORTION PLUS NOISE
vs
OUTPUT VOLTAGE

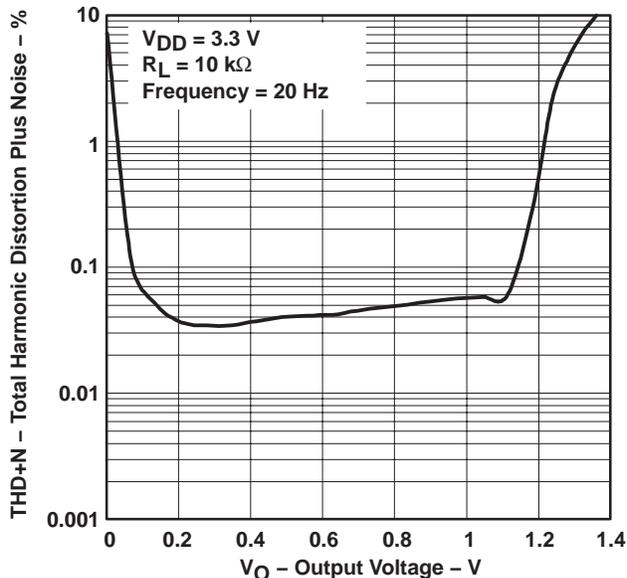


Figure 14

OUTPUT POWER
vs
LOAD RESISTANCE

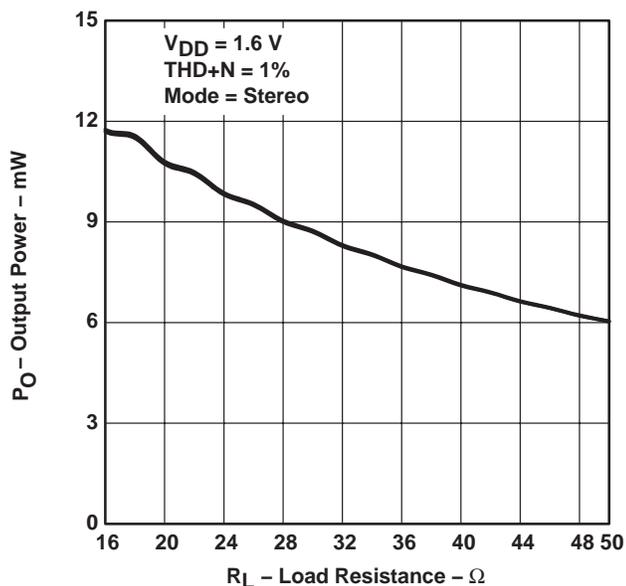


Figure 15

OUTPUT POWER
vs
LOAD RESISTANCE

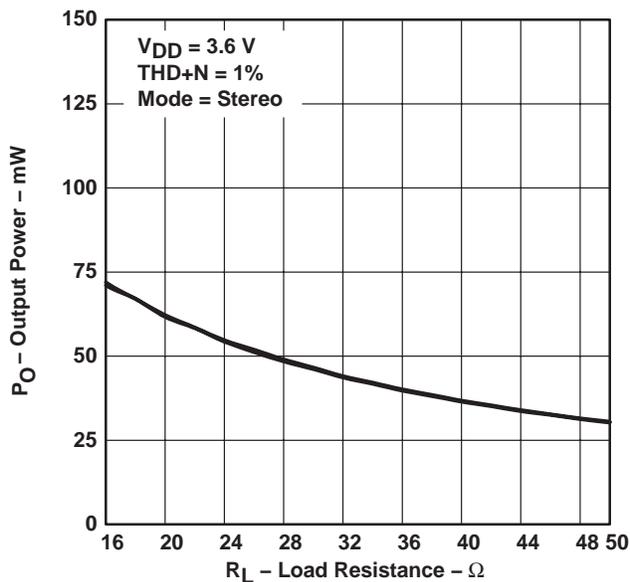


Figure 16

TPA6102A2
50-mW ULTRALOW-VOLTAGE, FIXED-GAIN STEREO HEADPHONE
AUDIO POWER AMPLIFIER

SLOS324B – JUNE 2000 – REVISED SEPTEMBER 2004

TYPICAL CHARACTERISTICS

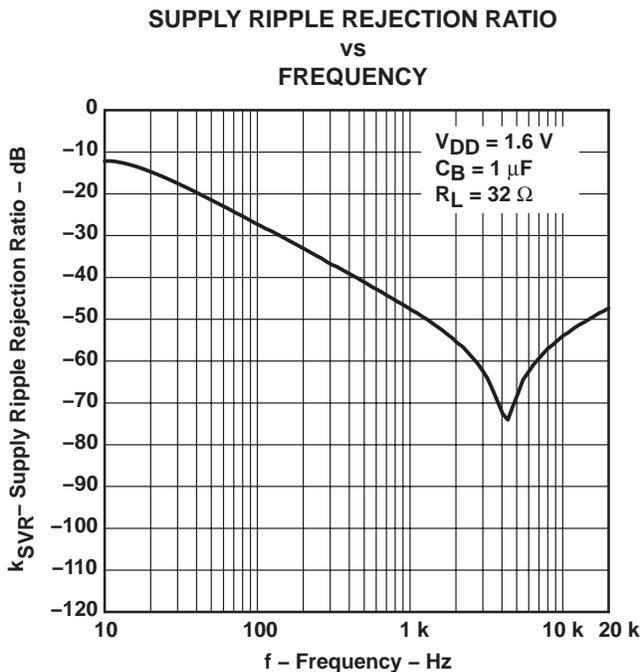


Figure 17

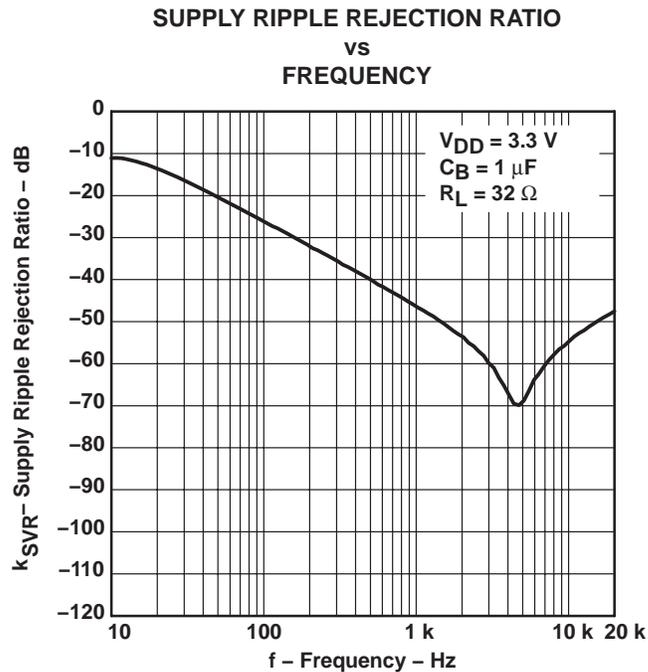


Figure 18

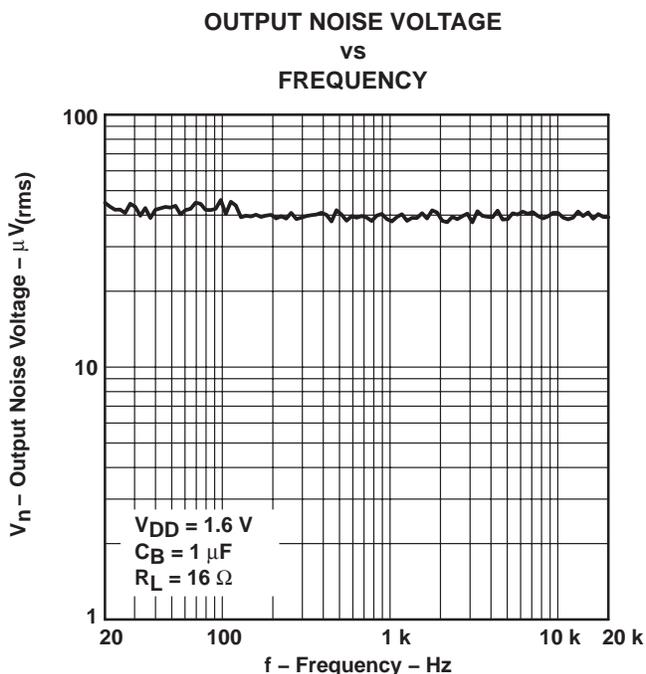


Figure 19

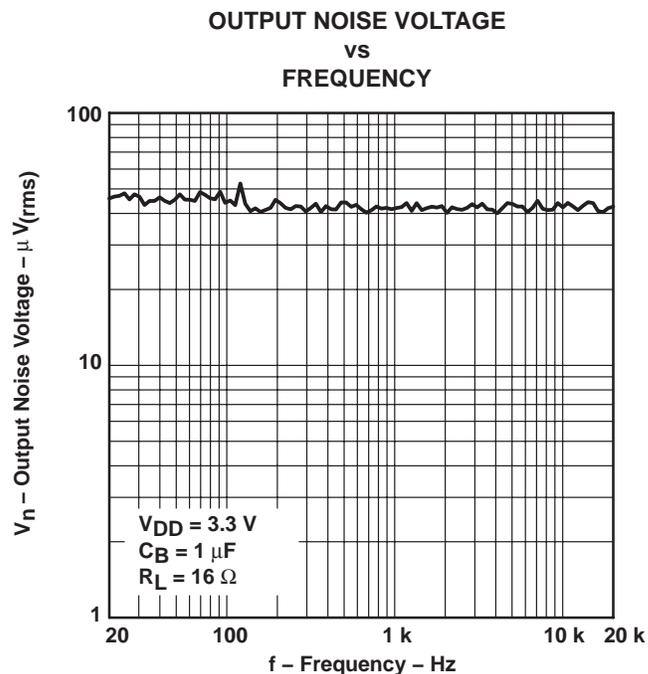


Figure 20

TYPICAL CHARACTERISTICS

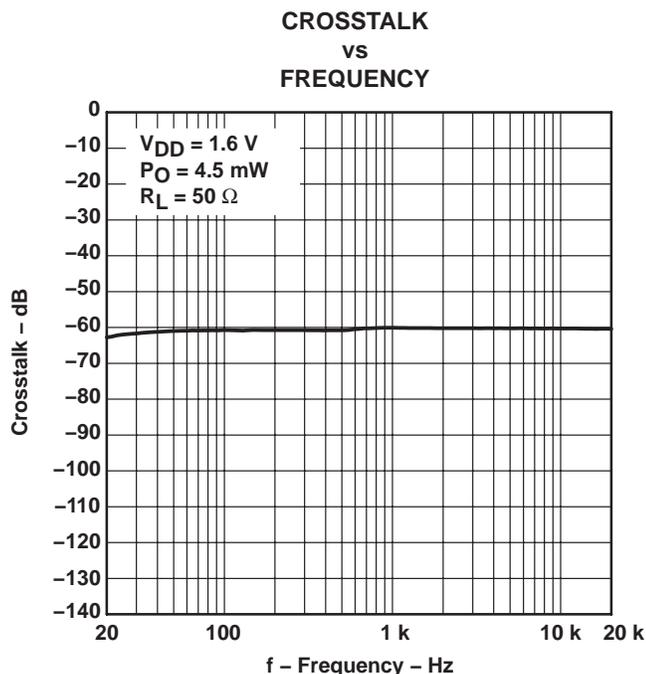


Figure 21

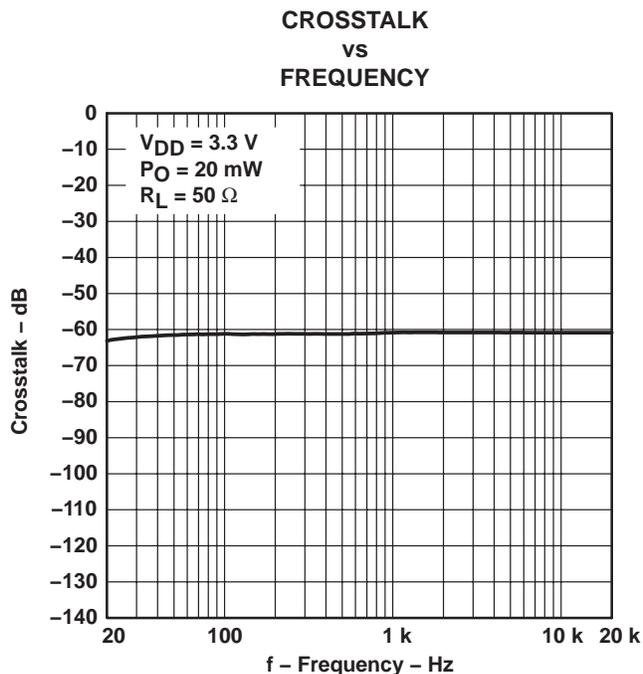


Figure 22

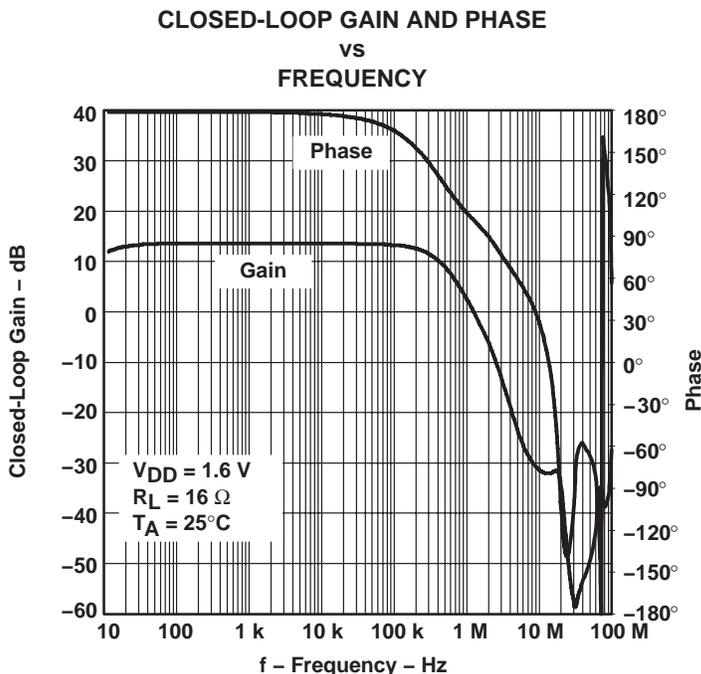


Figure 23

TYPICAL CHARACTERISTICS

**CLOSED-LOOP GAIN AND PHASE
 VS
 FREQUENCY**

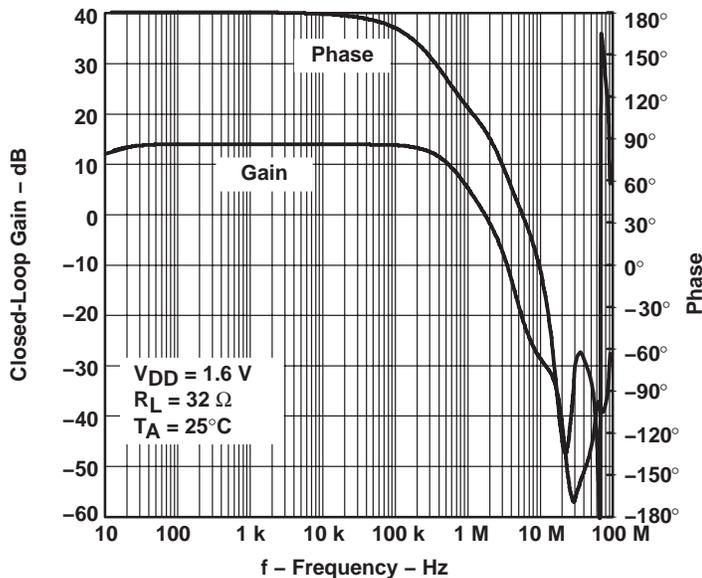


Figure 24

**CLOSED-LOOP GAIN AND PHASE
 VS
 FREQUENCY**

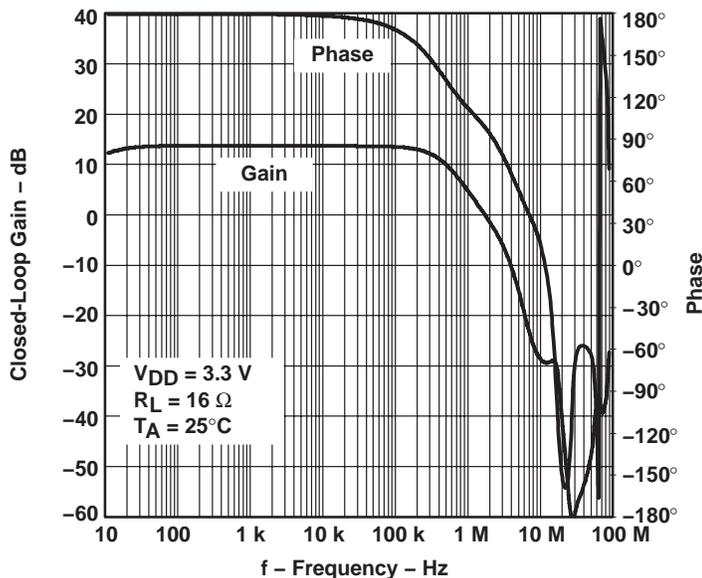


Figure 25

TYPICAL CHARACTERISTICS

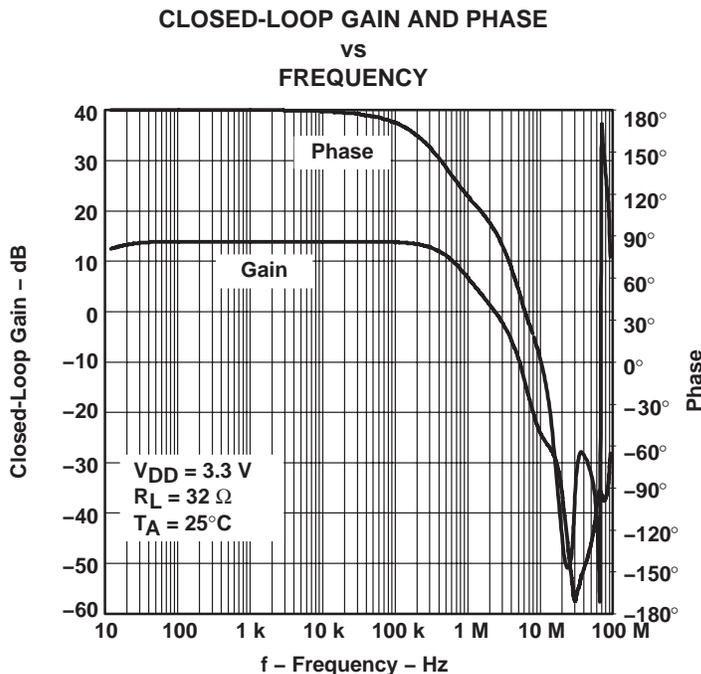


Figure 26

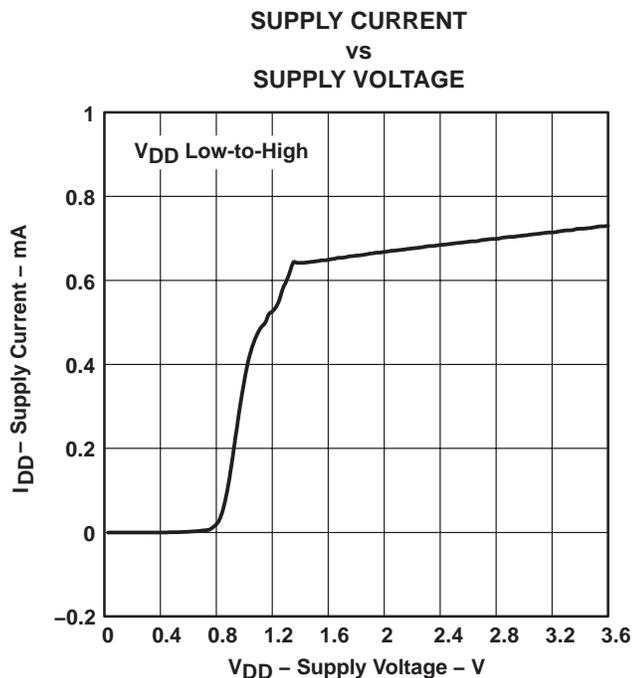


Figure 27

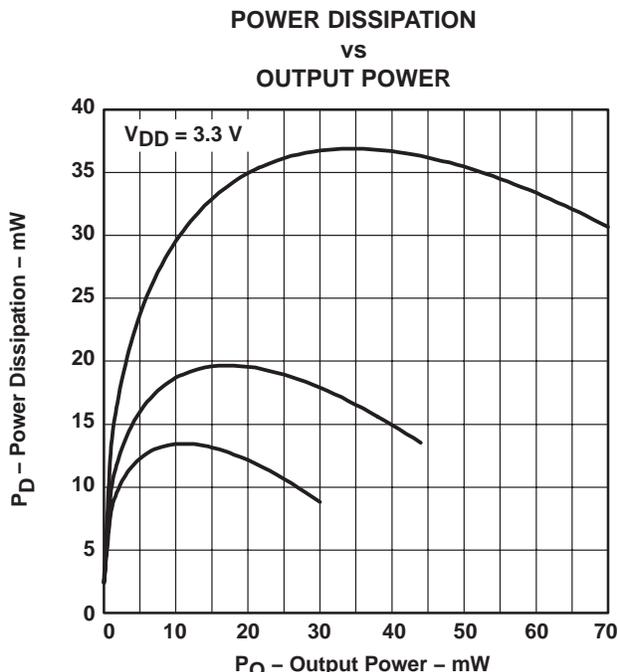


Figure 28

APPLICATION INFORMATION

input capacitor, C_I

In the typical application, an input capacitor (C_I) is required to allow the amplifier to bias the input signal to the proper dc level for optimum operation. In this case, C_I and R_I form a high-pass filter with the corner frequency determined in equation 1. R_I is set internally and is fixed at 20 k Ω .

$$f_c = \frac{1}{2\pi R_I C_I} \quad (1)$$

The value of C_I is important to consider, as it directly affects the bass (low frequency) performance of the circuit. Consider the example where the specification calls for a flat bass response down to 20 Hz. Equation 1 is reconfigured as equation 2.

$$C_I = \frac{1}{2\pi R_I f_c} \quad (2)$$

In this example, C_I is 0.40 μF , so one would likely choose a value in the range of 0.47 μF to 1 μF . A further consideration for this capacitor is the leakage path from the input source through the input network (R_I , C_I) and the feedback resistor (R_F) to the load. This leakage current creates a dc offset voltage at the input to the amplifier that reduces useful headroom. For this reason a low-leakage tantalum or ceramic capacitor is the best choice. When polarized capacitors are used, the positive side of the capacitor should face the amplifier input in most applications, as the dc level there is held at $V_{DD}/4$, which is likely higher than the source dc level. It is important to confirm the capacitor polarity in the application.

power supply decoupling, C_S

The TPA6102A2 is a high-performance CMOS audio amplifier that requires adequate power supply decoupling to ensure that the output total harmonic distortion (THD) is as low as possible. Power supply decoupling also prevents oscillations for long lead lengths between the amplifier and the speaker. The optimum decoupling is achieved by using two capacitors of different types that target different types of noise on the power supply leads. For higher frequency transients, spikes, or digital hash on the line, a good low equivalent-series-resistance (ESR) ceramic capacitor, typically 0.1 μF , placed as close as possible to the device V_{DD} lead, works best. For filtering lower-frequency noise signals, a larger aluminum electrolytic capacitor of 10 μF or greater placed near the power amplifier is recommended.

midrail bypass capacitor, C_B

The midrail bypass capacitor (C_B) serves several important functions. During start-up, C_B determines the rate at which the amplifier starts up. This helps to push the start-up pop noise into the subaudible range (so low it can not be heard). The second function is to reduce noise produced by the power supply caused by coupling into the output drive signal. This noise is from the midrail generation circuit internal to the amplifier. The capacitor is fed from a 55-k Ω source inside the amplifier. To keep the start-up pop as low as possible, the relationship shown in equation 3 should be maintained.

$$\frac{1}{(C_B \times 55 \text{ k}\Omega)} \leq \frac{1}{(C_I R_I)} \quad (3)$$

As an example, consider a circuit where C_B is 1 μF , C_I is 1 μF , and R_I is 20 k Ω . Inserting these values into the equation 3 results in: $18.18 \leq 50$ which satisfies the rule. Bypass capacitor (C_B) with values of 0.47- μF to 1- μF ceramic or tantalum low-ESR capacitors are recommended for the best THD and noise performance.

APPLICATION INFORMATION

output coupling capacitor, C_C

In the typical single-supply single-ended (SE) configuration, an output coupling capacitor (C_C) is required to block the dc bias at the output of the amplifier, thus preventing dc currents in the load. As with the input coupling capacitor, the output coupling capacitor and impedance of the load form a high-pass filter governed by equation 4.

$$f_c = \frac{1}{2\pi R_L C_C} \quad (4)$$

The main disadvantage, from a performance standpoint, is that the typically small load impedances drive the low-frequency corner higher. Large values of C_C are required to pass low-frequencies into the load. Consider the example where a C_C of 68 μF is chosen and loads vary from 32 Ω to 47 k Ω . Table 1 summarizes the frequency response characteristics of each configuration.

Table 1. Common-Load Impedances vs Low-Frequency Output Characteristics in SE Mode

R_L	C_C	Lowest Frequency
32 Ω	68 μF	73 Hz
10,000 Ω	68 μF	0.23 Hz
47,000 Ω	68 μF	0.05 Hz

As Table 1 indicates, headphone response is adequate and drive into line level inputs (a home stereo for example) is very good.

The output-coupling capacitor required in single-supply SE mode also places additional constraints on the selection of other components in the amplifier circuit. With the rules described earlier still valid, add the following relationship:

$$\frac{1}{(C_B \times 55 \text{ k}\Omega)} \leq \frac{1}{(C_I R_I)} \ll \frac{1}{R_L C_C} \quad (5)$$

using low-ESR capacitors

Low-ESR capacitors are recommended throughout this application. A real capacitor can be modeled simply as a resistor in series with an ideal capacitor. The voltage drop across this resistor minimizes the beneficial effects of the capacitor in the circuit. The lower the equivalent value of this resistance, the more the real capacitor behaves like an ideal capacitor.

3.3-V versus 1.6-V operation

The TPA6102A2 was designed for operation over a supply range of 1.6 V to 3.6 V. There are no special considerations for 1.6-V versus 3.3-V operation as far as supply bypassing, gain setting, or stability. Supply current is slightly reduced from 0.75 mA (typical) to 0.65 mA (typical). The most important consideration is that of output power. Each amplifier can produce a maximum output voltage swing within a few hundred millivolts of the rails with a 10-k Ω load. However, this voltage swing decreases as the load resistance decreases and the $r_{DS(on)}$ as the output stage transistors becomes more significant. For example, for a 32- Ω load, the maximum peak output voltage with $V_{DD} = 1.6$ V is approximately 0.7 V with no clipping distortion. This reduced voltage swing effectively reduces the maximum undistorted output power.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TPA6102A2D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	6102A2	Samples
TPA6102A2DG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	6102A2	Samples
TPA6102A2DGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	AJN	Samples
TPA6102A2DGKR	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	AJN	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

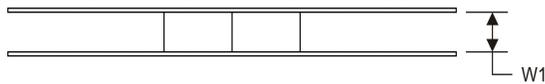
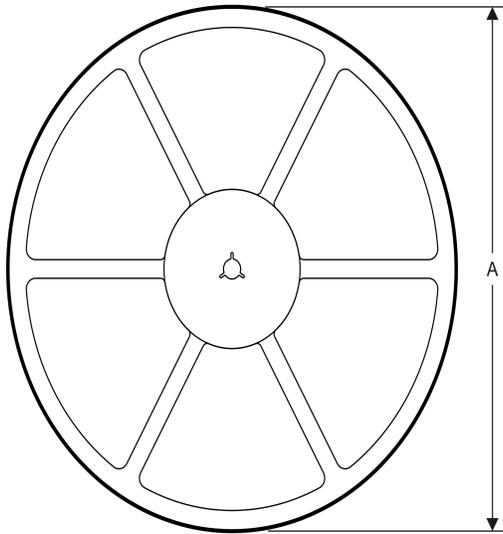
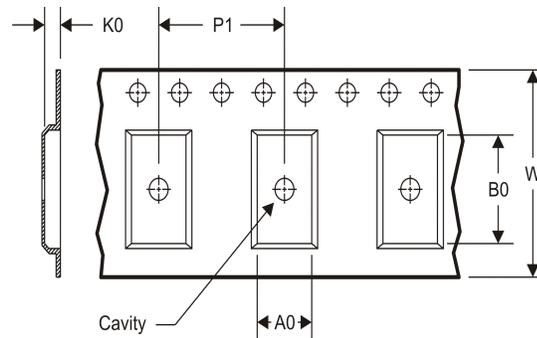
(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

TAPE AND REEL INFORMATION
REEL DIMENSIONS

TAPE DIMENSIONS


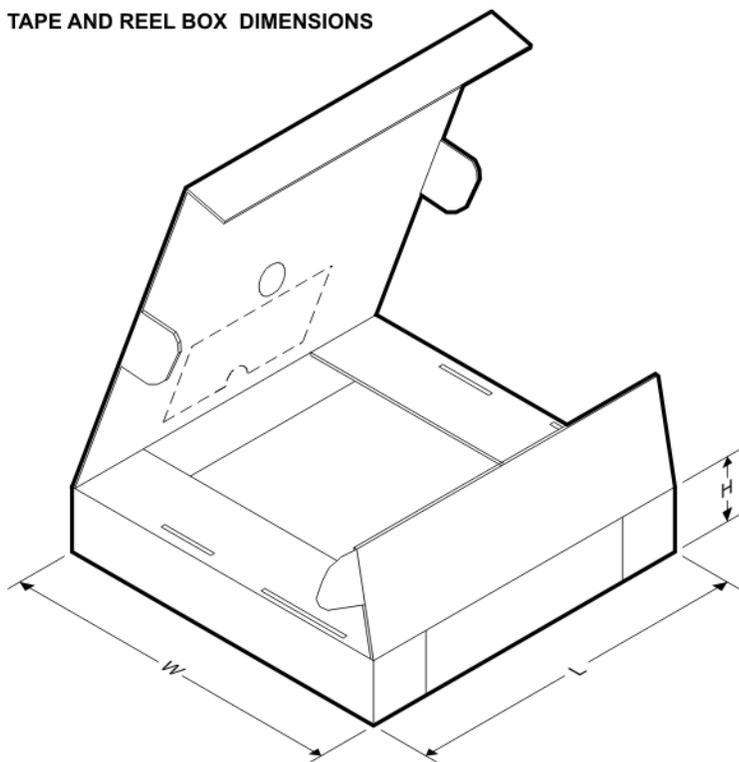
A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

TAPE AND REEL INFORMATION

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPA6102A2DGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS

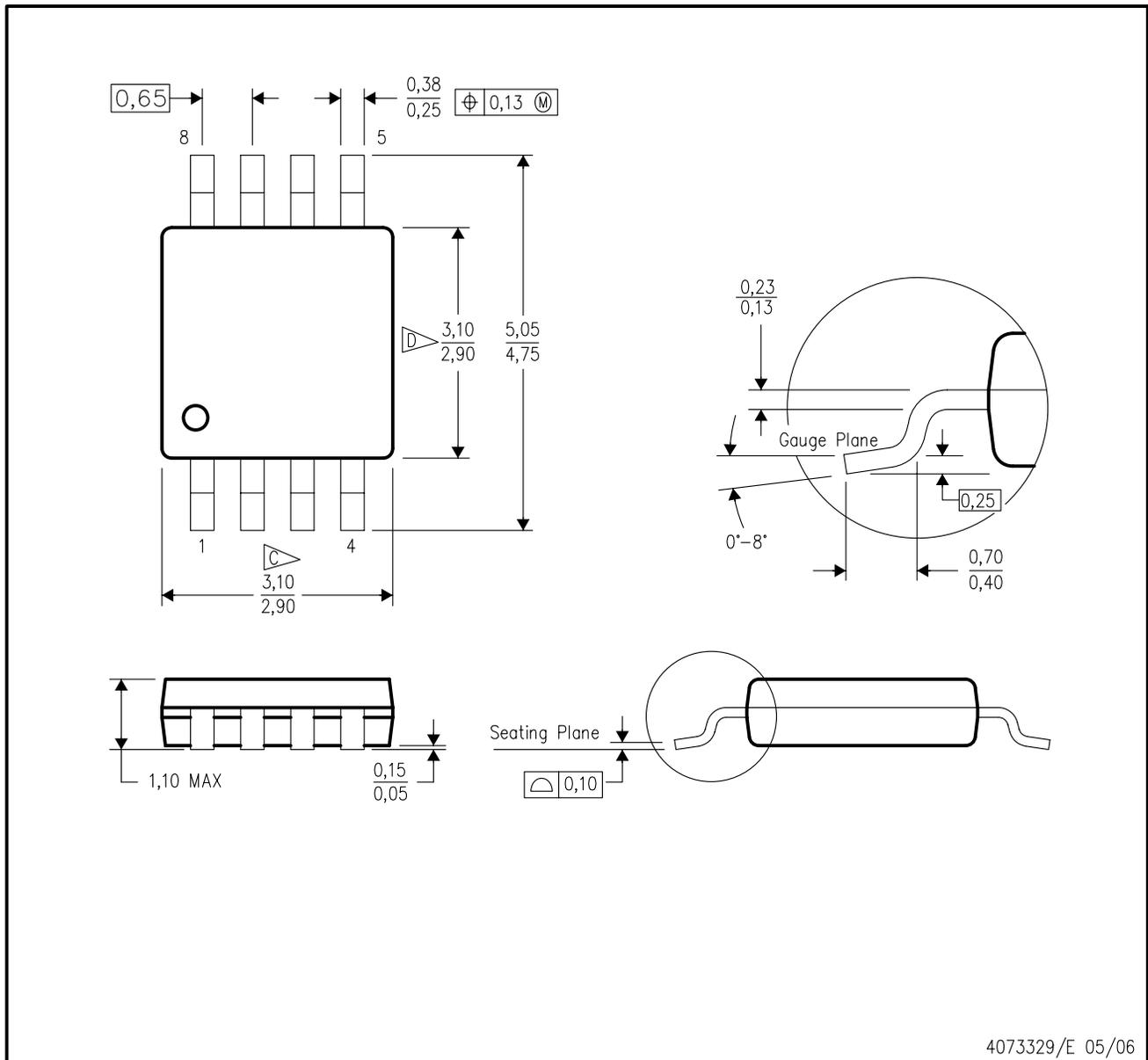


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPA6102A2DGKR	VSSOP	DGK	8	2500	358.0	335.0	35.0

DGK (S-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE

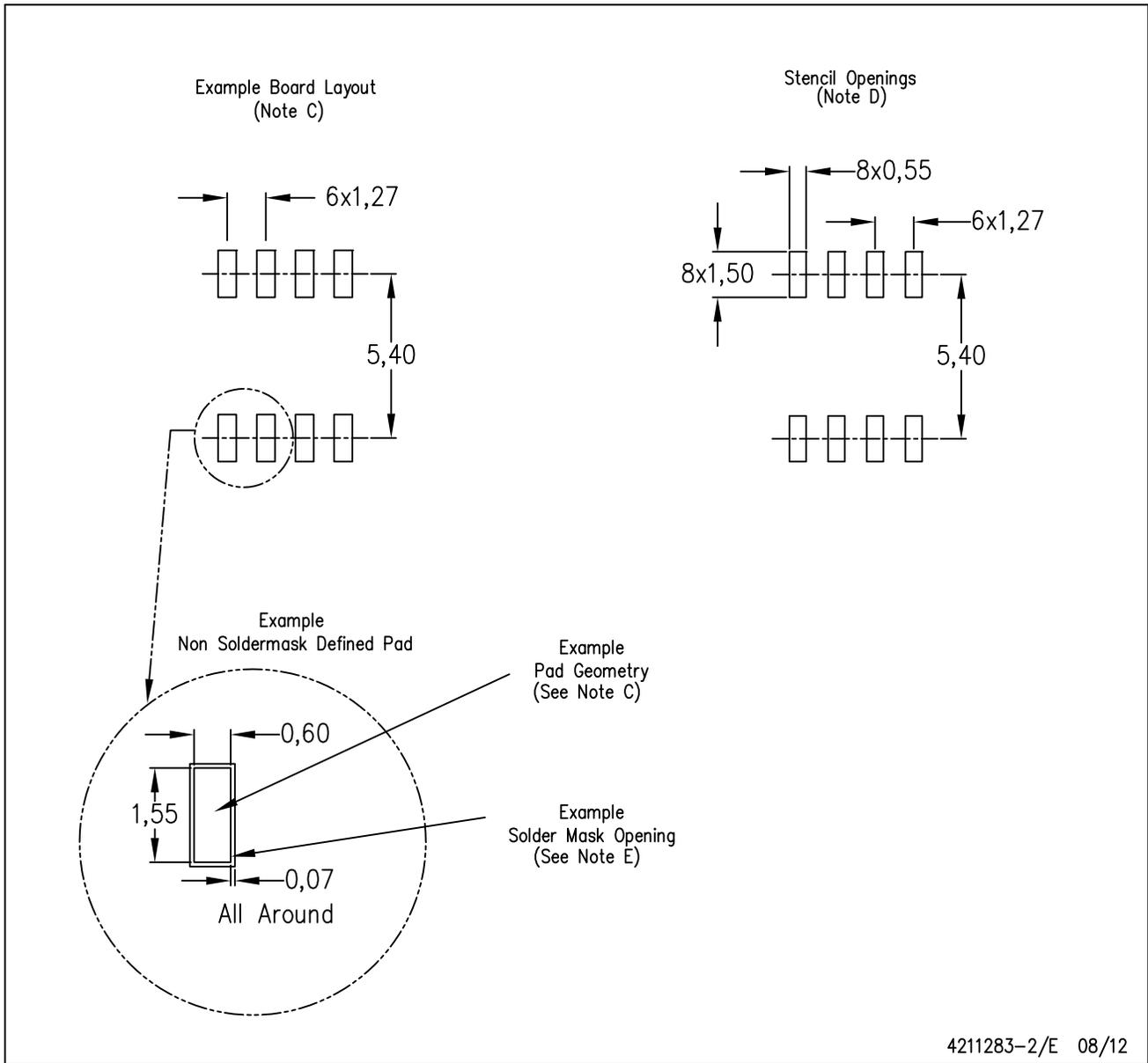


4073329/E 05/06

- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 per end.
 - D. Body width does not include interlead flash. Interlead flash shall not exceed 0.50 per side.
 - E. Falls within JEDEC MO-187 variation AA, except interlead flash.

D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Publication IPC-7351 is recommended for alternate designs.
 - D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
 - E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

IMPORTANT NOTICE

Texas Instruments Incorporated and its subsidiaries (TI) reserve the right to make corrections, enhancements, improvements and other changes to its semiconductor products and services per JESD46, latest issue, and to discontinue any product or service per JESD48, latest issue. Buyers should obtain the latest relevant information before placing orders and should verify that such information is current and complete. All semiconductor products (also referred to herein as "components") are sold subject to TI's terms and conditions of sale supplied at the time of order acknowledgment.

TI warrants performance of its components to the specifications applicable at the time of sale, in accordance with the warranty in TI's terms and conditions of sale of semiconductor products. Testing and other quality control techniques are used to the extent TI deems necessary to support this warranty. Except where mandated by applicable law, testing of all parameters of each component is not necessarily performed.

TI assumes no liability for applications assistance or the design of Buyers' products. Buyers are responsible for their products and applications using TI components. To minimize the risks associated with Buyers' products and applications, Buyers should provide adequate design and operating safeguards.

TI does not warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right relating to any combination, machine, or process in which TI components or services are used. Information published by TI regarding third-party products or services does not constitute a license to use such products or services or a warranty or endorsement thereof. Use of such information may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

Reproduction of significant portions of TI information in TI data books or data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. TI is not responsible or liable for such altered documentation. Information of third parties may be subject to additional restrictions.

Resale of TI components or services with statements different from or beyond the parameters stated by TI for that component or service voids all express and any implied warranties for the associated TI component or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

Buyer acknowledges and agrees that it is solely responsible for compliance with all legal, regulatory and safety-related requirements concerning its products, and any use of TI components in its applications, notwithstanding any applications-related information or support that may be provided by TI. Buyer represents and agrees that it has all the necessary expertise to create and implement safeguards which anticipate dangerous consequences of failures, monitor failures and their consequences, lessen the likelihood of failures that might cause harm and take appropriate remedial actions. Buyer will fully indemnify TI and its representatives against any damages arising out of the use of any TI components in safety-critical applications.

In some cases, TI components may be promoted specifically to facilitate safety-related applications. With such components, TI's goal is to help enable customers to design and create their own end-product solutions that meet applicable functional safety standards and requirements. Nonetheless, such components are subject to these terms.

No TI components are authorized for use in FDA Class III (or similar life-critical medical equipment) unless authorized officers of the parties have executed a special agreement specifically governing such use.

Only those TI components which TI has specifically designated as military grade or "enhanced plastic" are designed and intended for use in military/aerospace applications or environments. Buyer acknowledges and agrees that any military or aerospace use of TI components which have **not** been so designated is solely at the Buyer's risk, and that Buyer is solely responsible for compliance with all legal and regulatory requirements in connection with such use.

TI has specifically designated certain components as meeting ISO/TS16949 requirements, mainly for automotive use. In any case of use of non-designated products, TI will not be responsible for any failure to meet ISO/TS16949.

Products

Audio	www.ti.com/audio
Amplifiers	amplifier.ti.com
Data Converters	dataconverter.ti.com
DLP® Products	www.dlp.com
DSP	dsp.ti.com
Clocks and Timers	www.ti.com/clocks
Interface	interface.ti.com
Logic	logic.ti.com
Power Mgmt	power.ti.com
Microcontrollers	microcontroller.ti.com
RFID	www.ti-rfid.com
OMAP Applications Processors	www.ti.com/omap
Wireless Connectivity	www.ti.com/wirelessconnectivity

Applications

Automotive and Transportation	www.ti.com/automotive
Communications and Telecom	www.ti.com/communications
Computers and Peripherals	www.ti.com/computers
Consumer Electronics	www.ti.com/consumer-apps
Energy and Lighting	www.ti.com/energy
Industrial	www.ti.com/industrial
Medical	www.ti.com/medical
Security	www.ti.com/security
Space, Avionics and Defense	www.ti.com/space-avionics-defense
Video and Imaging	www.ti.com/video

TI E2E Community

e2e.ti.com