

Precision 4.096V SOT-23 FGA™ Voltage References

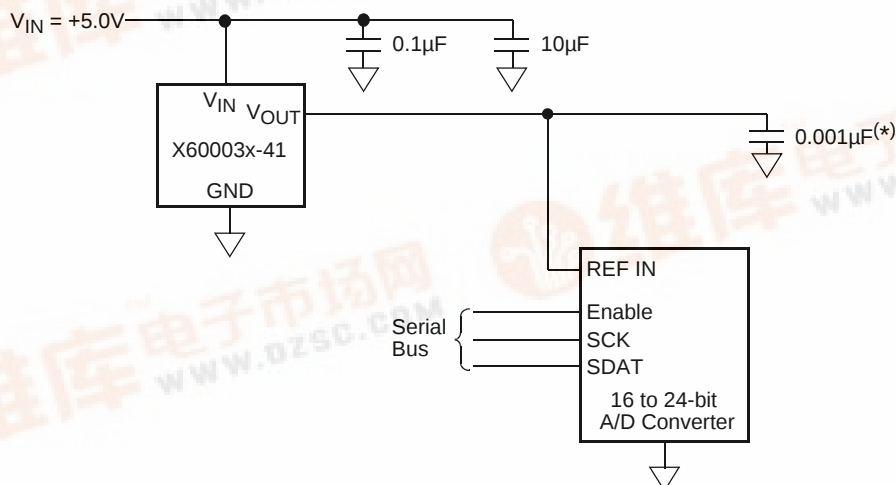
FEATURES

- Output Voltage: 4.096V
- Absolute Initial Accuracy Options: $\pm 1.0\text{mV}$, $\pm 2.5\text{mV}$, & $\pm 5.0\text{mV}$
- Ultra Low Power Supply Current: 500nA
- Low Temperature Coefficient Options: 10 & 20ppm/°C
- 10 mA Source & Sink Current Capability
- 10 ppm/1000hrs Long Term Stability
- Supply Voltage Range: 4.5V to 9.0V
- 5kV ESD (Human Body Model)
- Standard Package: 3-lead SOT-23
- Temp Range: -40°C to +85°C

APPLICATIONS

- High Resolution A/Ds & D/As
- Digital Meters
- Calibration Systems
- V-F Converters
- Precision Current Sources
- Precision Regulators
- Precision Oscillators
- Battery Management Systems
- Smart sensors
- Strain Gage Bridges
- Threshold Detectors
- Servo Systems

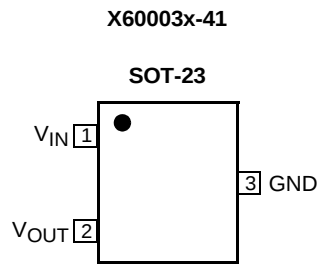
TYPICAL APPLICATION



(*)Also see Figure 3 in Applications Information

X6003B-41, X6003C-41, X6003D-41

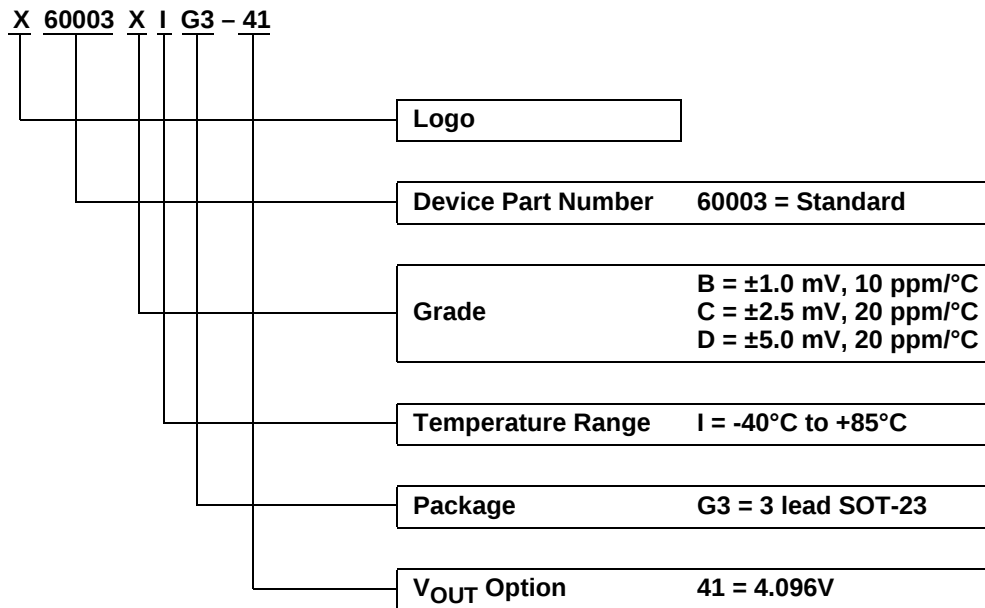
PACKAGE DIAGRAM



PIN CONFIGURATIONS

Pin Name	Description
GND	Ground Connection
V _{IN}	Power Supply Input Connection
V _{OUT}	Voltage Reference Output Connection

ORDERING INFORMATION



PART MARKING

Order Part Number	Top Marking
X60003BIG3-41	AHA
X60003CIG3-41	AHB
X60003DIG3-41	AHC

X6003B-41, X6003C-41, X6003D-41

ABSOLUTE MAXIMUM RATINGS

Storage Temperature Range..... -65°C to + 125°C
 Max Voltage Applied
 V_{IN} to Gnd.....-0.5V to + 10V
 Max Voltage Applied
 V_{OUT} to Gnd(*).....-0.5V to + 5.1V
 Lead Temperature (soldering, 10 secs)..... + 225°C

*Maximum duration = 10 seconds

RECOMMENDED OPERATING CONDITIONS

Temperature	Min.	Max.
Industrial	-40°C	+85°C

COMMENT

Absolute Maximum Ratings indicate limits beyond which permanent damage to the device and impaired reliability may occur. These are stress ratings provided for information only and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of this specification are not implied.

For guaranteed specifications and test conditions, see Electrical Characteristics.

The guaranteed specifications apply only for the test conditions listed. Some performance characteristics may degrade when the device is not operated under the listed test conditions.

ELECTRICAL CHARACTERISTICS

(Operating Conditions: $V_{IN} = 5.0V$, $I_{OUT} = 0mA$, $C_{OUT} = 0.001\mu F$, $T_A = -40$ to $+85^\circ C$ unless otherwise specified.)

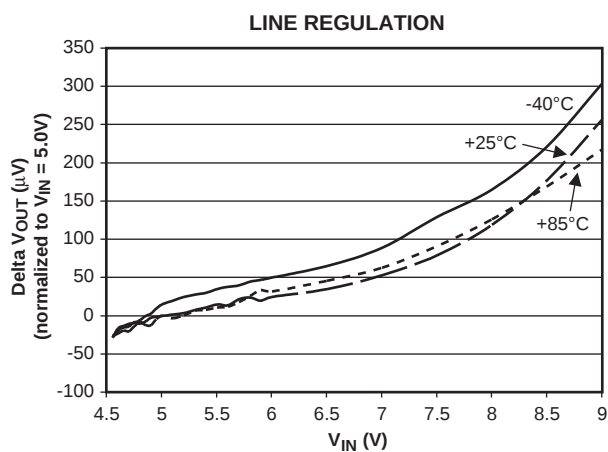
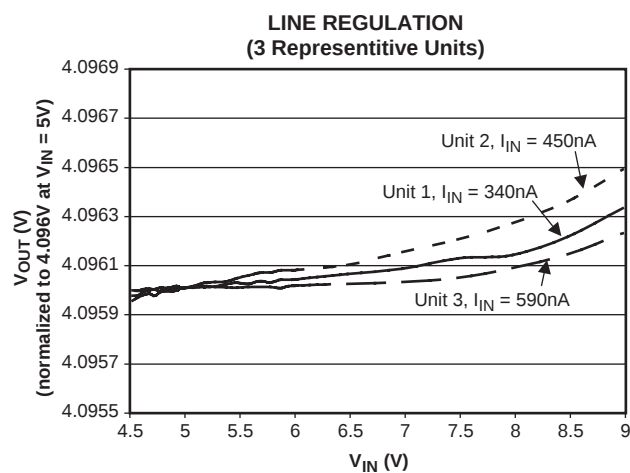
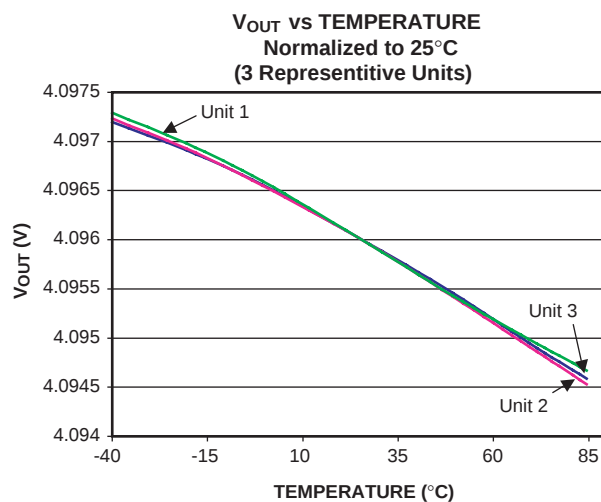
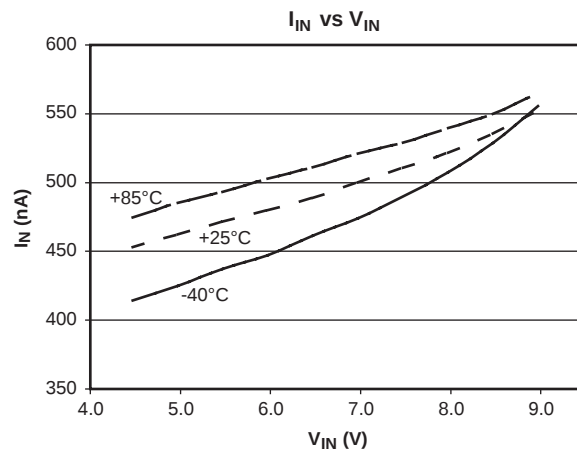
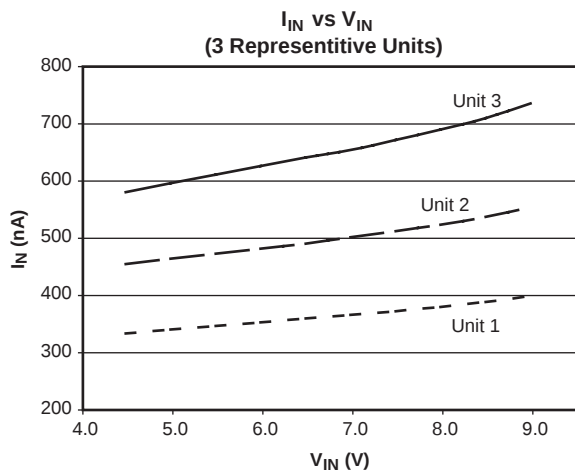
Symbol	Parameter	Conditions	Min	Typ	Max	Units
V_{OUT}	Output Voltage			4.096		V
V_{OA}	V_{OUT} Accuracy X60003B-41 X60003C-41 X60003D-41	$T_A = 25^\circ C$	-1.0 -2.5 -5.0		+1.0 +2.5 +5.0	mV
I_{IN}	Supply Current			500	900	nA
V_{IN}	Input Voltage Range		4.5		9.0	V
TC V_{OUT}	Output Voltage Temperature Coefficient ⁽¹⁾	X60003B-41 X60003C-41 X60003D-41			10 20 20	ppm/ $^\circ C$
$\Delta V_{OUT}/\Delta V_{IN}$	Line Regulation	$+4.5V \leq V_{IN} \leq +8.0V$			150	$\mu V/V$
$\Delta V_{OUT}/\Delta I_{OUT}$	Load Regulation	$0mA \leq I_{SOURCE} \leq 10mA$ $-10mA \leq I_{SINK} \leq 0mA$		10 20	50 100	$\mu V/mA$
$\Delta V_{OUT}/\Delta t$	Long Term Stability	$T_A = 25^\circ C$		10		ppm/1000Hrs
$\Delta V_{OUT}/\Delta T_A$	Thermal Hysteresis ⁽²⁾	$\Delta T = -40^\circ C$ to $+85^\circ C$		150		ppm
I_{SC}	Short Circuit Current ⁽³⁾	$T_A = 25^\circ C$		50	80	mA
V_N	Output Voltage Noise	0.1Hz to 10Hz		30		μV_{pp}

- Note: 1. Over the specified temperature range. Temperature coefficient is measured by the box method whereby the change in V_{OUT} is divided by the temperature range; in this case, $-40^\circ C$ to $+85^\circ C = 125^\circ C$.
 2. Thermal Hysteresis is the change in V_{OUT} created by package stress @ $T_A = 25^\circ C$ after temperature cycling. V_{OUT} is read initially at $T_A = 25^\circ C$; the X60003x-41 is then cycled between Hot ($85^\circ C$) and Cold ($-40^\circ C$) before a second V_{OUT} measurement is taken at $25^\circ C$. The deviation between the initial V_{OUT} reading and the second V_{OUT} reading is then expressed in ppm.
 3. Guaranteed by Device Characterization and/or correlation to other device tests.

X6003B-41, X6003C-41, X6003D-41

TYPICAL PERFORMANCE CHARACTERISTIC CURVES

($V_{IN} = 5.0V$, $I_{OUT} = 0mA$, $T_A = 25^\circ C$ unless otherwise specified)

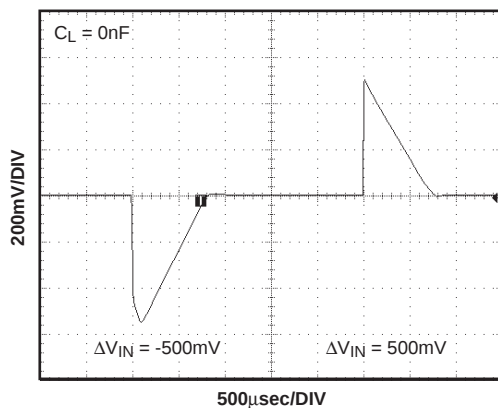


X6003B-41, X6003C-41, X6003D-41

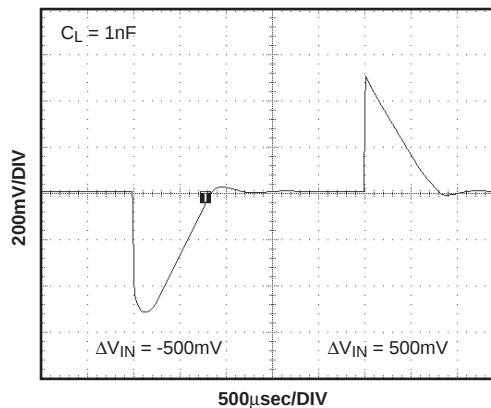
TYPICAL PERFORMANCE CHARACTERISTIC CURVES

($V_{IN} = 5.0V$, $I_{OUT} = 0mA$, $T_A = 25^\circ C$ unless otherwise specified)

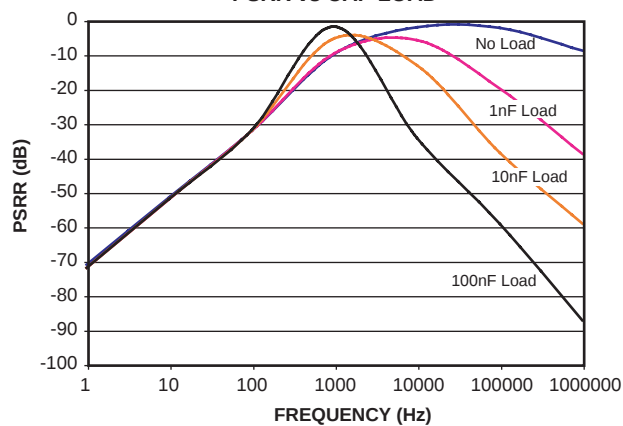
LINE TRANSIENT RESPONSE



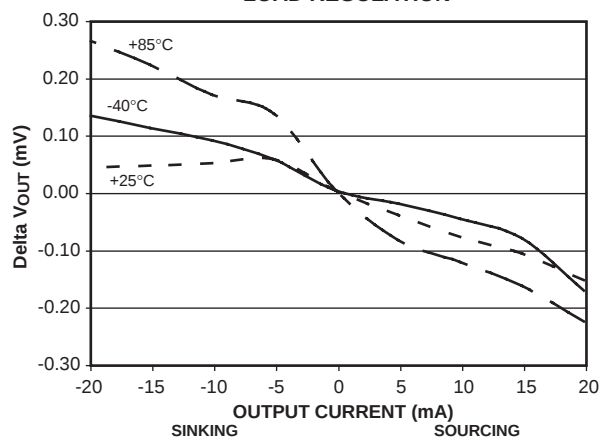
LINE TRANSIENT RESPONSE



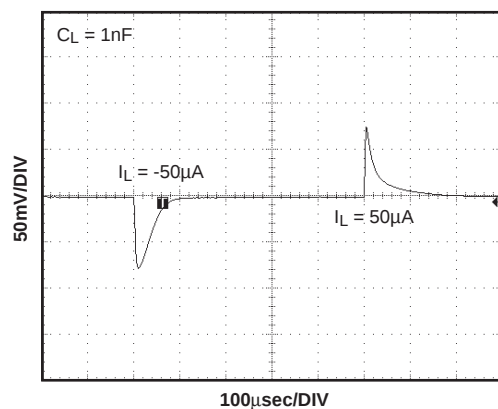
PSRR vs CAP LOAD



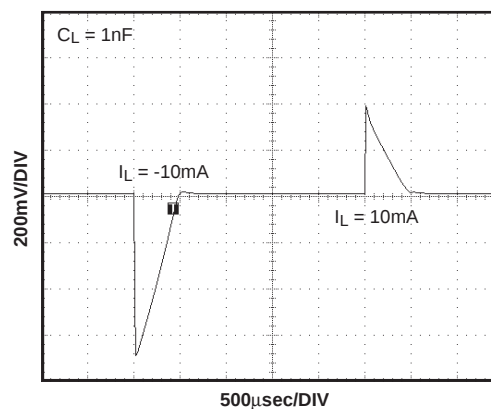
LOAD REGULATION



LOAD TRANSIENT RESPONSE



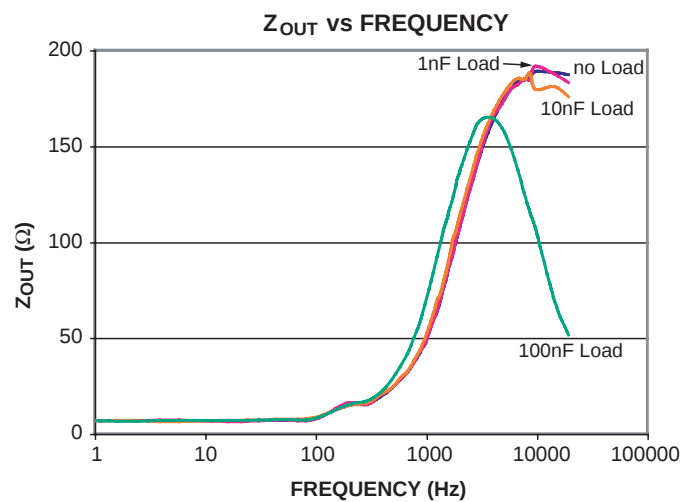
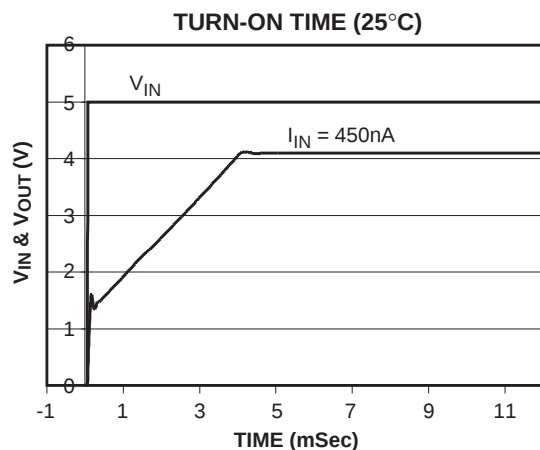
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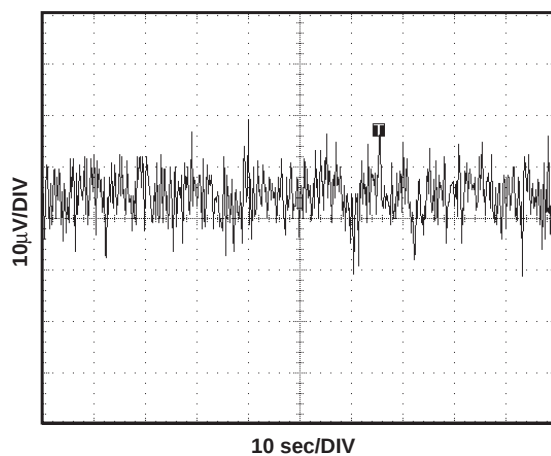
X6003B-41, X6003C-41, X6003D-41

TYPICAL PERFORMANCE CHARACTERISTIC CURVES

($V_{IN} = 5.0V$, $I_{OUT} = 0mA$, $T_A = 25^\circ C$ unless otherwise specified)



0.1Hz to 10Hz V_{OUT} NOISE
Band Pass Filter with 1 Zero at 0.1Hz and 2 Poles at 10Hz



APPLICATIONS INFORMATION

FGA Technology

The X60003x-41 voltage reference uses the floating gate technology to create references with very low drift and supply current. Essentially the charge stored on a floating gate cell is set precisely in manufacturing. The reference voltage output itself is a buffered version of the floating gate voltage. The resulting reference device has excellent characteristics which are unique in the industry: very low temperature drift, high initial accuracy, and almost zero supply current. Also, the reference voltage itself is not limited by voltage bandgaps or zener settings, so a wide range of reference voltages can be programmed (standard voltage settings are provided, but customer-specific voltages are available).

The process used for these reference devices is a floating gate CMOS process, and the amplifier circuitry uses CMOS transistors for amplifier and output transistor circuitry. While providing excellent accuracy, there are limitations in output noise level and load regulation due to the MOS device characteristics. These limitations are addressed with circuit techniques discussed in other sections.

Nanopower Operation

Reference devices achieve their highest accuracy when powered up continuously, and after initial stabilization has taken place.

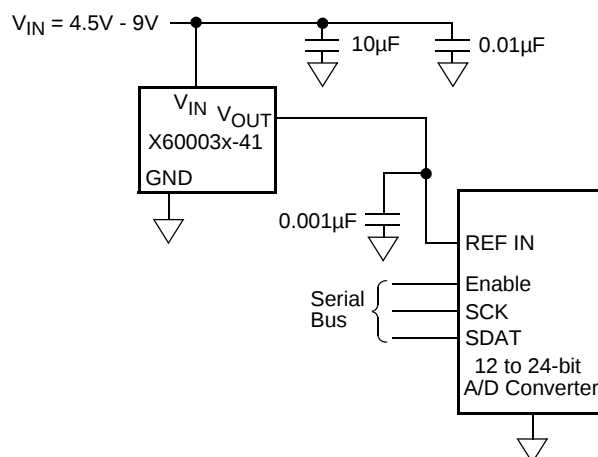
The X60003x-41 is the first high precision voltage reference with ultra low power consumption that makes it practical to leave power-on continuously in battery operated circuits. The X60003x-41 consumes extremely low supply current due to the proprietary FGA technology. Supply current at room temperature is typically 500nA which is 1 to 2 orders of magnitude lower than competitive devices. Application circuits using battery power will benefit greatly from having an accurate, stable reference which essentially presents no load to the battery.

In particular, battery powered data converter circuits that would normally require the entire circuit to be disabled when not in use can remain powered up between conversions as shown in figure 1. Data acquisition circuits providing 12 to 24 bits of accuracy can operate with the reference device continuously biased with no power penalty, providing the highest accuracy and lowest possible long term drift.

Other reference devices consuming higher supply currents will need to be disabled in between conversions to conserve battery capacity. Absolute accuracy will

suffer as the device is biased and requires time to settle to its final value, or, may not actually settle to a final value as power-on time may be short.

Figure 1.



Board mounting Considerations

For applications requiring the highest accuracy, board mounting location should be reviewed. Placing the device in areas subject to slight twisting can cause degradation of the accuracy of the reference voltage due to die stresses. It is normally best to place the device near the edge of a board, or the shortest side, as the axis of bending is most limited at that location. Obviously mounting the device on flexprint or extremely thin PC material will likewise cause loss of reference accuracy.

Noise Performance and Reduction:

The output noise voltage in a 0.1Hz to 10Hz bandwidth is typically 30 μV p-p. This is shown in the plot in the Typical Performance Curves. The noise measurement is made with a bandpass filter made of a 1 pole high-pass filter with a corner frequency at .1Hz and a 2-pole low-pass filter with a corner frequency at 12.6Hz to create a filter with a 9.9Hz bandwidth. Noise in the 10KHz to 1MHz bandwidth is approximately 400 μV p-p with no capacitance on the output, as shown in Fig. 2 below. These noise measurements are made with a 2 decade bandpass filter made of a 1 pole high-pass filter with a corner frequency at 1/10 of the center frequency and 1-pole low-pass filter with a corner frequency at 10 times the center frequency. Figure 2 also shows the noise in the 10KHz to 1MHz band can be reduced to about 50 μV p-p using a .001 μF capacitor on the output. Noise in the 1KHz to 100KHz band can be further reduced using a 0.1 μF capacitor on the output, but noise in the 1Hz to 100Hz band increases due to instability of the very low power amplifier with a 0.1 μF capacitance load. For

X6003B-41, X6003C-41, X6003D-41

load capacitances above $.001\mu\text{F}$ the noise reduction network shown in fig. 3 is recommended. This network reduces noise significantly over the full bandwidth. As shown in fig. 2, noise is reduced to less than $40\mu\text{Vp-p}$ from 1Hz to 1MHz using this network with a $.01\mu\text{F}$ capacitor and a $2\text{k}\Omega$ resistor in series with a $10\mu\text{F}$ capacitor.

Figure 2.

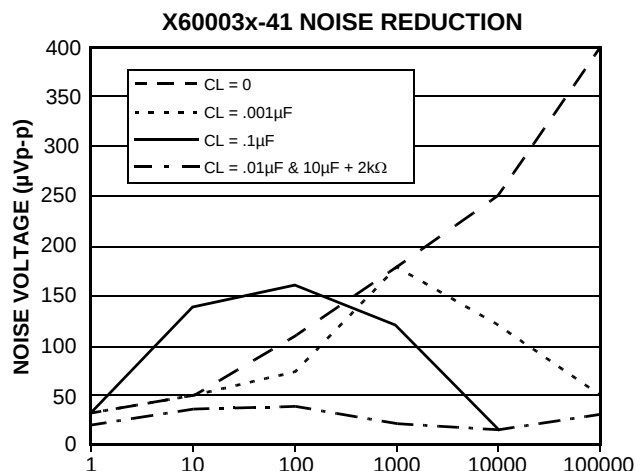
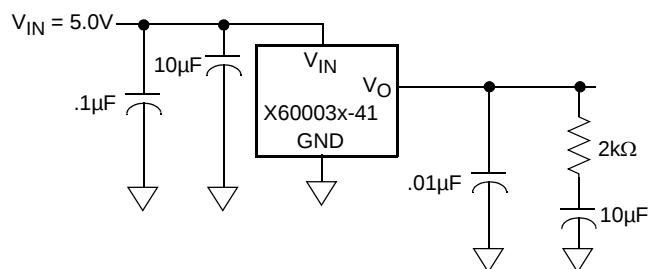


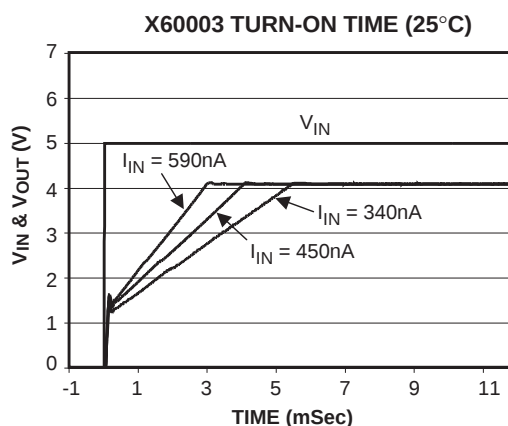
Figure 3.



Turn-On Time

The X60003x-41 device has ultra-low supply current and thus the time to bias up internal circuitry to final values will be longer than with higher power references. Normal turn-on time is typically 7ms. This is shown in the graph, Figure 4. Since devices can vary in supply current down to 300nA, turn-on time can last up to about 12ms. Care should be taken in system design to include this delay before measurements or conversions are started.

Figure 4.



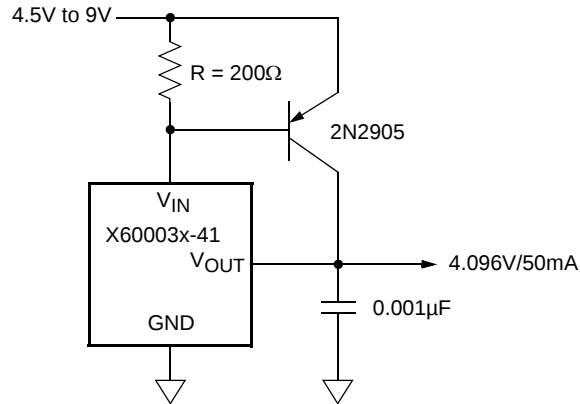
Temperature Coefficient

The limits stated for temperature coefficient (tempco) are governed by the method of measurement. The overwhelming standard for specifying the temperature drift of a reference is to measure the reference voltage at two temperatures, take the total variation, $(V_{\text{HIGH}} - V_{\text{LOW}})$, and divide by the temperature extremes of measurement $(T_{\text{HIGH}} - T_{\text{LOW}})$. The result is divided by the nominal reference voltage (at $T = 25^\circ\text{C}$) and multiplied by 10^6 to yield ppm/ $^\circ\text{C}$. This is the "Box" method for determining temperature coefficient.

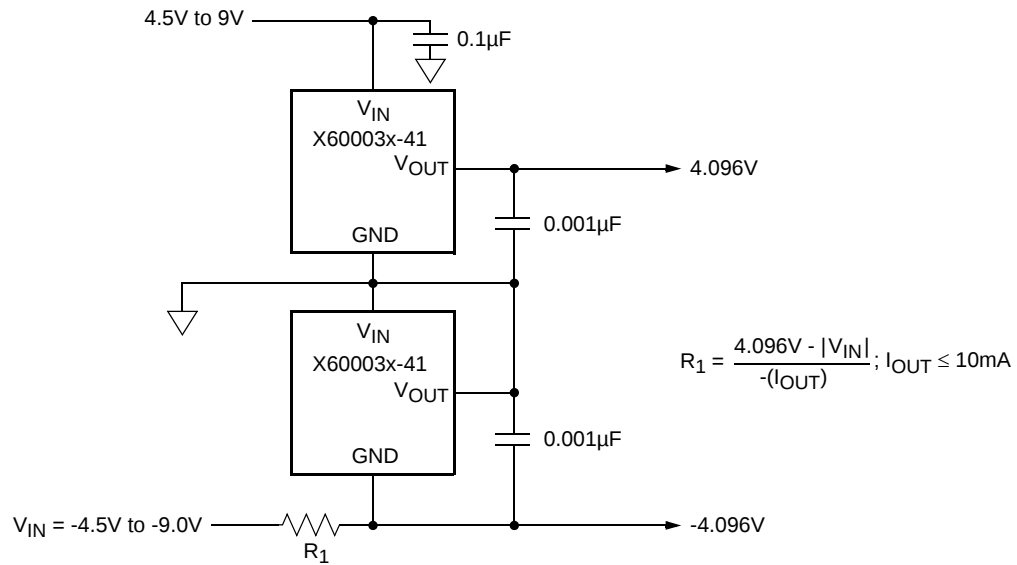
X6003B-41, X6003C-41, X6003D-41

TYPICAL APPLICATION CIRCUITS

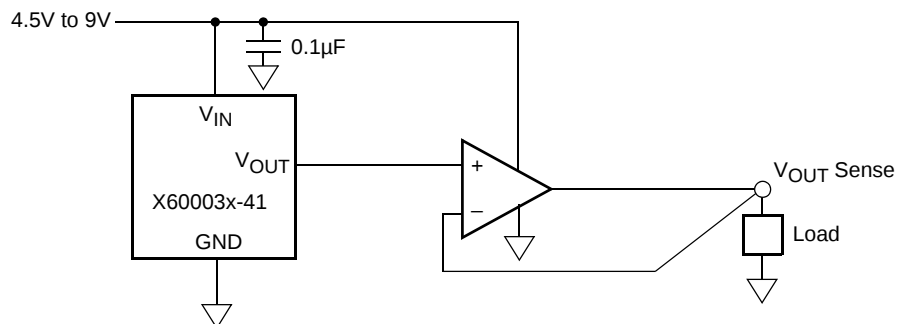
Precision 4.096V, 50mA Reference.



± 4.096 V Dual Output, High Accuracy Reference



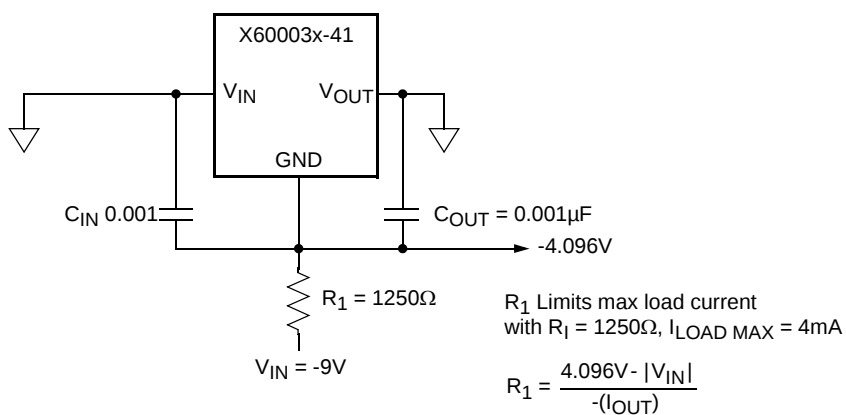
Kelvin Sensed Load



X6003B-41, X6003C-41, X6003D-41

TYPICAL APPLICATION CIRCUITS

Negative Voltage Reference



4.096V Full Scale Low-Drift 10-bit Adjustable Voltage Source

