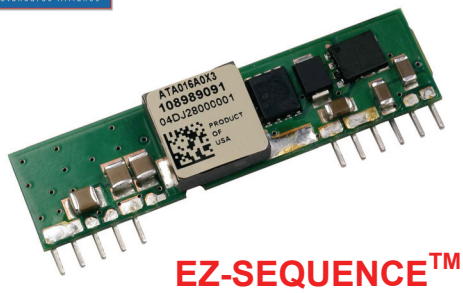
Data Sheet
April 1, 2008

LINEAGE POWER

Austin SuperLynx™ II 12V SIP Non-isolated Power Modules: 8.3Vdc – 14Vdc input; 0.75Vdc to 5.5Vdc Output; 16A Output Current

DOSA Distributed-power Open Standards Alliance

RoHS Compliant



EZ-SEQUENCE™

Applications

- Distributed power architectures
- Intermediate bus voltage applications
- Telecommunications equipment
- Servers and storage applications
- Networking equipment
- Enterprise Networks
- Latest generation IC's (DSP, FPGA, ASIC) and Microprocessor powered applications

Features

- Compliant to RoHS EU Directive 2002/95/EC (-Z versions)
- Compliant to ROHS EU Directive 2002/95/EC with lead solder exemption (non-Z versions)
- Flexible output voltage sequencing EZ-SEQUENCE™
- Delivers up to 16A output current
- High efficiency – 92% at 3.3V full load (V_{IN} = 12.0V)
- Small size and low profile:

  50.8 mm x 12.7 mm x 8.1 mm

  (2.00 in x 0.5 in x 0.32 in)
- Low output ripple and noise
- Constant switching frequency (300KHz)
- High Reliability:

  Calculated MTBF = 9.2M hours at 25°C Full-load
- Programmable Output voltage
- Line Regulation: 0.3% (typical)
- Load Regulation: 0.4% (typical)
- Temperature Regulation: 0.4 % (typical)
- Remote On/Off
- Remote Sense
- Output overcurrent protection (non-latching)
- Wide operating temperature range (-40°C to 85°C)
- *UL** 60950-1Recognized, *CSA*† C22.2 No. 60950-1-03 Certified, and *VDE*‡ 0805:2001-12 (EN60950-1) Licensed
- ISO** 9001 and ISO 14001 certified manufacturing facilities

Description

Austin SuperLynx™ II 12V SIP (single in-line package) power modules are non-isolated dc-dc converters that can deliver up to 16A of output current with full load efficiency of 92% at 3.3V output. These modules provide a precisely regulated output voltage programmable via an external resistor from 0.75Vdc to 5.0Vdc over a wide range of input voltage (V_{IN} = 8.3 – 14Vdc). Austin SuperLynx™ II has a sequencing feature, EZ-SEQUENCE™ that enable designers to implement various types of output voltage sequencing when powering multiple modules on board. Their open-frame construction and small footprint enable designers to develop cost- and space-efficient solutions.

* *UL* is a registered trademark of Underwriters Laboratories, Inc.
† *CSA* is a registered trademark of Canadian Standards Association.
‡ *VDE* is a trademark of Verband Deutscher Elektrotechniker e.V.
** ISO is a registered trademark of the International Organization of Standards

Absolute Maximum Ratings

Stresses in excess of the absolute maximum ratings can cause permanent damage to the device. These are absolute stress ratings only, functional operation of the device is not implied at these or any other conditions in excess of those given in the operations sections of the data sheet. Exposure to absolute maximum ratings for extended periods can adversely affect the device reliability.

Parameter	Device	Symbol	Min	Max	Unit
Input Voltage Continuous	All	V_{IN}	-0.3	15	Vdc
Sequencing voltage	All	Vseq	-0.3	$V_{IN,max}$	Vdc
Operating Ambient Temperature (see Thermal Considerations section)	All	T_A	-40	85	°C
Storage Temperature	All	T_{stg}	-55	125	°C

Electrical Specifications

Unless otherwise indicated, specifications apply over all operating input voltage, resistive load, and temperature conditions.

Parameter	Device	Symbol	Min	Typ	Max	Unit
Operating Input Voltage	$V_{o,set} \leq 3.63$ $V_{o,set} > 3.63$	V_{IN} V_{IN}	8.3 8.3	12.0 12.0	14.0 13.2	Vdc Vdc
Maximum Input Current (V_{IN}= $V_{IN, min}$ to $V_{IN, max}$, I_O=$I_{O, max}$)	All	$I_{IN,max}$			10	Adc
Input No Load Current (V_{IN} = $V_{IN, nom}$, Io = 0, module enabled)	Vo = 0.75Vdc Vo = 5.0Vdc	$I_{IN,No\ load}$ $I_{IN,No\ load}$		40 100		mA mA
Input Stand-by Current (V_{IN} = $V_{IN, nom}$, module disabled)	All	$I_{IN,stand-by}$		2		mA
Inrush Transient	All	I^2t			0.4	A^2s
Input Reflected Ripple Current, peak-to-peak (5Hz to 20MHz, 1µH source impedance; V_{IN}=10V to 14V, I_O= I_{Omax} ; See Test configuration section)	All			30		mAp-p
Input Ripple Rejection (120Hz)	All			30		dB

CAUTION: This power module is not internally fused. An input line fuse must always be used.

This power module can be used in a wide variety of applications, ranging from simple standalone operation to being part of a complex power architecture. To preserve maximum flexibility, internal fusing is not included, however, to achieve maximum safety and system protection, always use an input line fuse. The safety agencies require a fast-acting fuse with a maximum rating of 15 A (see Safety Considerations section). Based on the information provided in this data sheet on inrush energy and maximum dc input current, the same type of fuse with a lower rating can be used. Refer to the fuse manufacturer's data sheet for further information.

Electrical Specifications (continued)

Parameter	Device	Symbol	Min	Typ	Max	Unit
Output Voltage Set-point (V_{IN}=$V_{IN, min}$, I_O=$I_{O, max}$, T_A=25°C)	All	$V_{O, set}$	-2.0	$V_{O, set}$	+2.0	% $V_{O, set}$
Output Voltage (Over all operating input voltage, resistive load, and temperature conditions until end of life)	All	$V_{O, set}$	-2.5%	—	+3.5%	% $V_{O, set}$
Adjustment Range Selected by an external resistor	All	V_O	0.7525		5.5	Vdc
Output Regulation						
Line (V_{IN}=$V_{IN, min}$ to $V_{IN, max}$)	All		—	0.3	—	% $V_{O, set}$
Load (I_O=$I_{O, min}$ to $I_{O, max}$)	All		—	0.4	—	% $V_{O, set}$
Temperature (T_{ref}=$T_{A, min}$ to $T_{A, max}$)	All		—	0.4	—	% $V_{O, set}$
Output Ripple and Noise on nominal output (V_{IN}=$V_{IN, nom}$ and I_O=$I_{O, min}$ to $I_{O, max}$ Cout = 1µF ceramic//10µFtantalum capacitors)						
RMS (5Hz to 20MHz bandwidth)	Vo ≤ 3.63		—	12	30	mV_{rms}
Peak-to-Peak (5Hz to 20MHz bandwidth)	Vo ≤ 3.63		—	30	75	mV_{pk-pk}
RMS (5Hz to 20MHz bandwidth)	V_o = 5.0V		—	25	40	mV_{rms}
Peak-to-Peak (5Hz to 20MHz bandwidth)	V_o = 5.0V		—	70	100	mV_{pk-pk}
External Capacitance						
ESR ≥ 1 mΩ	All	$C_{O, max}$	—	—	1000	µF
ESR ≥ 10 mΩ	All	$C_{O, max}$	—	—	5000	µF
Output Current	All	I_o	0		16	Adc
Output Current Limit Inception (Hiccup Mode) (V_O= 90% of $V_{O, set}$)	All	$I_{O, lim}$	—	180	—	% I_o
Output Short-Circuit Current (V_O≤250mV) (Hiccup Mode)	All	$I_{O, s/c}$	—	3	—	Adc
Efficiency	$V_{O, set}$ = 0.75Vdc	η		79.0		%
V_{IN}= $V_{IN, nom}$, T_A=25°C	$V_{O, set}$ = 1.2Vdc	η		85.0		%
I_O=$I_{O, max}$, V_O= $V_{O,set}$	$V_{O,set}$ = 1.5Vdc	η		87.0		%
	$V_{O,set}$ = 1.8Vdc	η		88.0		%
	$V_{O,set}$ = 2.5Vdc	η		90.5		%
	$V_{O,set}$ = 3.3Vdc	η		92.0		%
	$V_{O,set}$ = 5.0Vdc	η		94.0		%
Switching Frequency	All	f_{sw}	—	300	—	kHz
Dynamic Load Response						
(dIo/dt=2.5A/µs; V_{IN} = $V_{IN, nom}$; T_A=25°C) Load Change from Io= 50% to 100% of Io,max; 1µF ceramic// 10 µF tantalum Peak Deviation	All	V_{pk}	—	200	—	mV
Settling Time (Vo<10% peak deviation)	All	t_s	—	25	—	µs
(dIo/dt=2.5A/µs; V_{IN} = $V_{IN, nom}$; T_A=25°C) Load Change from Io= 100% to 50%of Io,max: 1µF ceramic// 10 µF tantalum Peak Deviation	All	V_{pk}	—	200	—	mV
Settling Time (Vo<10% peak deviation)	All	t_s	—	25	—	µs

Electrical Specifications (continued)

Parameter	Device	Symbol	Min	Typ	Max	Unit
Dynamic Load Response						
(dIo/dt=2.5A/µs; V V_{IN} = $V_{IN, nom}$; T_A=25°C)	All	V_{pk}	—	100	—	mV
Load Change from Io= 50% to 100% of Io,max; Co = 2x150 µF polymer capacitors						
Peak Deviation						
Settling Time (Vo<10% peak deviation)	All	t_s	—	50	—	µs
(dIo/dt=2.5A/µs; V_{IN} = $V_{IN, nom}$; T_A=25°C)	All	V_{pk}	—	100	—	mV
Load Change from Io= 100% to 50%of Io,max: Co = 2x150 µF polymer capacitors						
Peak Deviation						
Settling Time (Vo<10% peak deviation)	All	t_s	—	50	—	µs

General Specifications

Parameter	Min	Typ	Max	Unit
Calculated MTBF (I_O=$I_{O, max}$, T_A=25°C) Telecordia SR-332 Issue 1: Method 1 Case 3		9,230,550		Hours
Weight	—	5.6 (0.2)	—	g (oz.)

Feature Specifications

Unless otherwise indicated, specifications apply over all operating input voltage, resistive load, and temperature conditions. See Feature Descriptions for additional information.

Parameter	Device	Symbol	Min	Typ	Max	Unit
On/Off Signal interface						
Device code with Suffix "4" – Positive logic						
(On/Off is open collector/drain logic input; Signal referenced to GND - See feature description section)						
Input High Voltage (Module ON)	All	V_{IH}	—	—	$V_{IN, max}$	V
Input High Current	All	I_{IH}	—	—	10	µA
Input Low Voltage (Module OFF)	All	V_{IL}	-0.2	—	0.3	V
Input Low Current	All	I_{IL}	—	0.2	1	mA
Device Code with no suffix – Negative Logic						
(On/OFF pin is open collector/drain logic input with external pull-up resistor; signal referenced to GND)						
Input High Voltage (Module OFF)	All	V_{IH}	2.5	—	$V_{IN,max}$	Vdc
Input High Current	All	I_{IH}		0.2	1	mA
Input Low Voltage (Module ON)	All	V_{IL}	-0.2	—	0.3	Vdc
Input low Current	All	I_{IL}		—	10	µA
Turn-On Delay and Rise Times						
(I_O=$I_{O, max}$, V_{IN} = $V_{IN, nom}$, T_A = 25 °C,)						
Case 1: On/Off input is set to Logic Low (Module ON) and then input power is applied (delay from instant at which V_{IN} =$V_{IN, min}$ until Vo=10% of Vo,set)	All	Tdelay	—	3	—	msec
Case 2: Input power is applied for at least one second and then the On/Off input is set to logic Low (delay from instant at which Von/Off=0.3V until Vo=10% of Vo, set)	All	Tdelay	—	3	—	msec
Output voltage Rise time (time for Vo to rise from 10% of $V_{o,set}$ to 90% of Vo, set)	All	Trise	—	4	6	msec
Output voltage overshoot – Startup				—	1	% $V_{O, set}$
I_O= $I_{O, max}$; V_{IN} = 8.3 to 14Vdc, T_A = 25 °C						
Sequencing Delay time						
Delay from $V_{IN, min}$ to application of voltage on SEQ pin	All	$T_{SEQ-delay}$	10			msec
Tracking Accuracy (Power-Up: 2V/ms)	All	$\lvert V_{SEQ} - V_o \rvert$		100	200	mV
(Power-Down: 1V/ms)	All	$\lvert V_{SEQ} - V_o \rvert$		300	500	mV
($V_{IN, min}$ to $V_{IN, max}$; $I_{O, min}$ to $I_{O, max}$ $V_{SEQ} < V_o$)						
Overtemperature Protection	All	T_{ref}	—	125	—	°C
(See Thermal Consideration section)						
Input Undervoltage Lockout						
Turn-on Threshold	All			7.9		V
Turn-off Threshold	All			7.8		V

Characteristic Curves

The following figures provide typical characteristics for the Austin SuperLynx™ II 12V SIP modules at 25ºC.

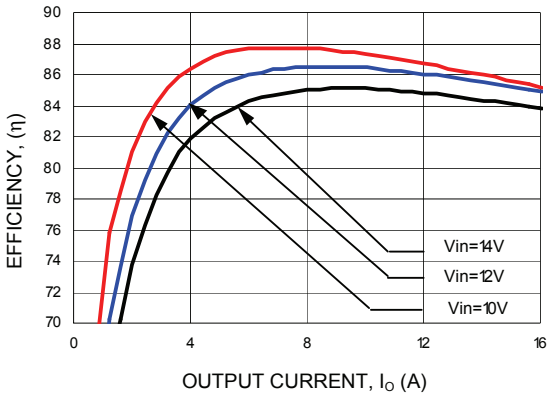


Figure 1. Converter Efficiency versus Output Current (Vout = 1.2Vdc).

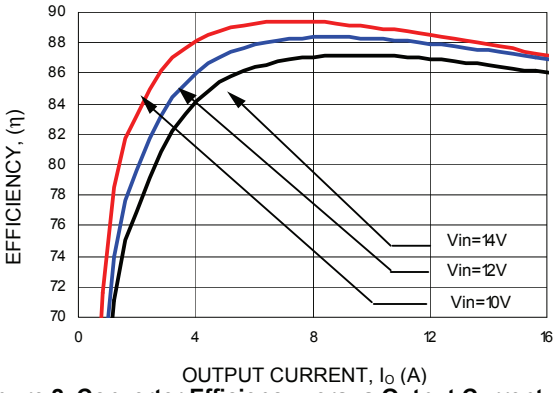


Figure 2. Converter Efficiency versus Output Current (Vout = 1.5Vdc).

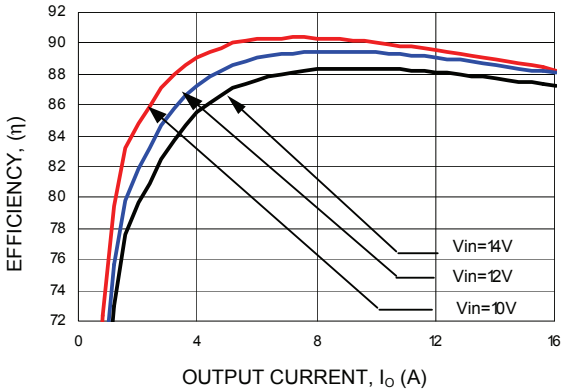


Figure 3. Converter Efficiency versus Output Current (Vout = 1.8Vdc).

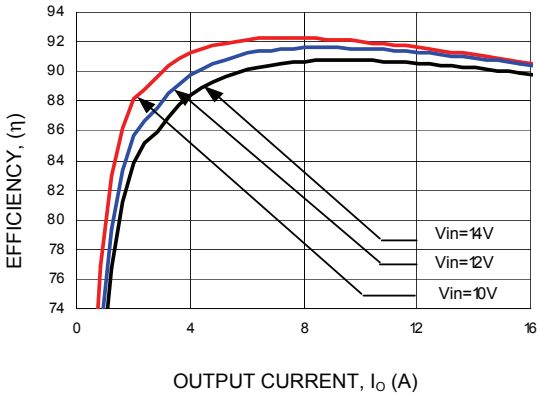


Figure 4. Converter Efficiency versus Output Current (Vout = 2.5Vdc).

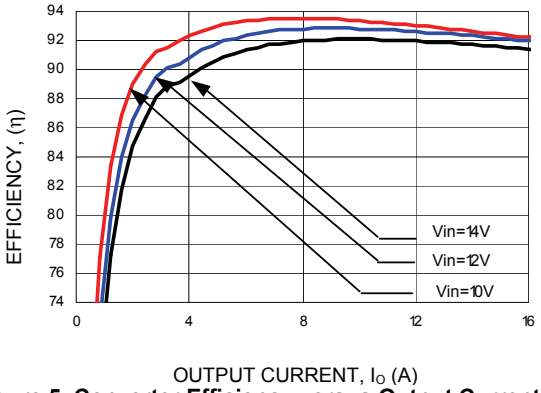


Figure 5. Converter Efficiency versus Output Current (Vout = 3.3Vdc).

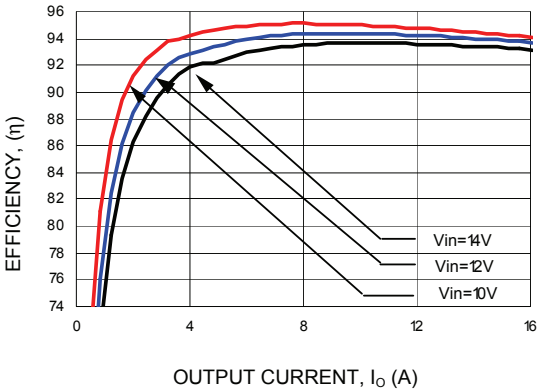


Figure 6. Converter Efficiency versus Output Current (Vout =5.0Vdc).

Characteristic Curves (continued)

The following figures provide typical characteristics for the SuperLynx™ II 12V SIP modules at 25ºC.

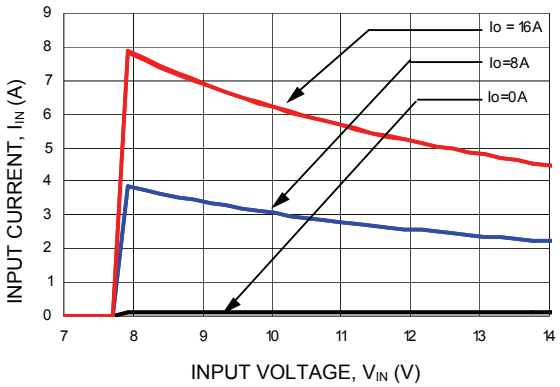


Figure 7. Input Voltage vs. Input Current (Vo = 3.3 Vdc).

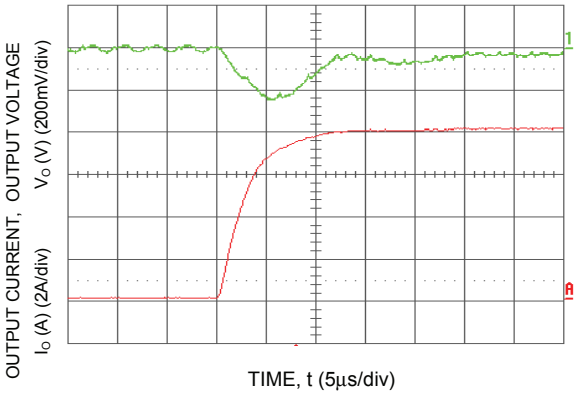


Figure 10. Transient Response to Dynamic Load Change from 50% to 100% of full load (Vo = 3.3Vdc).

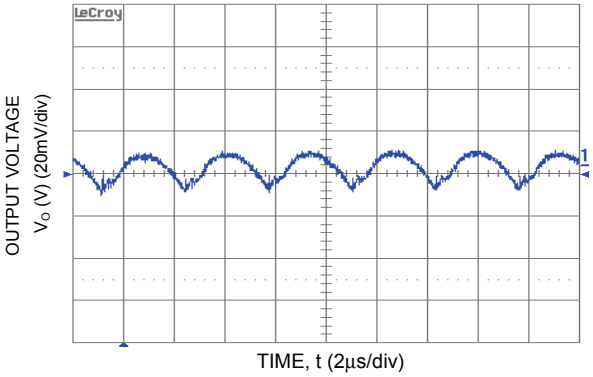


Figure 8. Typical Output Ripple and Noise (Vin = 12V dc, Vo = 2.5 Vdc, Io=16A).

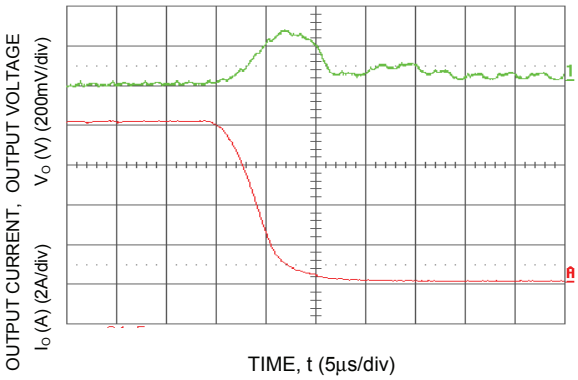


Figure 11. Transient Response to Dynamic Load Change from 100% to 50% of full load (Vo = 3.3Vdc).

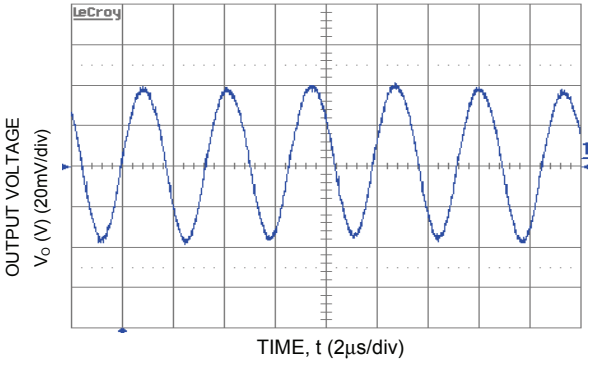


Figure 9. Typical Output Ripple and Noise (Vin = 12V dc, Vo = 3.3Vdc, Io=16A).

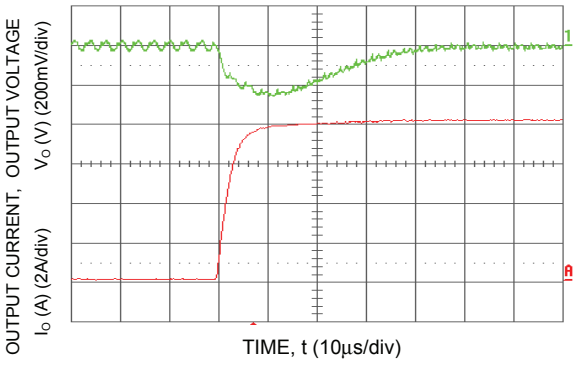


Figure 12. Transient Response to Dynamic Load Change from 50% to 100% of full load (Vo =3.3Vdc, Cext = 2x150 µF Polymer Capacitors).

Characteristic Curves (continued)

The following figures provide typical characteristics for the Austin SuperLynx™ II 12V SIP modules at 25ºC.

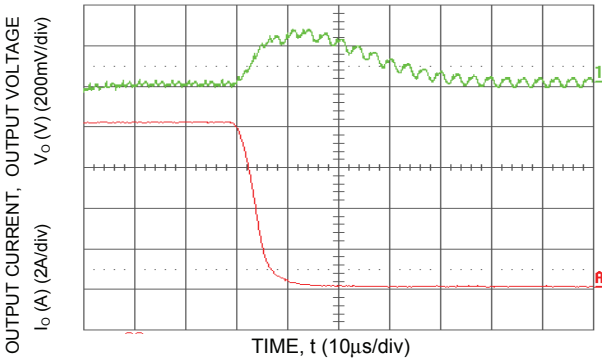


Figure 13. Transient Response to Dynamic Load Change from 100% of 50% full load (Vo = 3.3Vdc, Cext = 2x150 µF Polymer Capacitors)

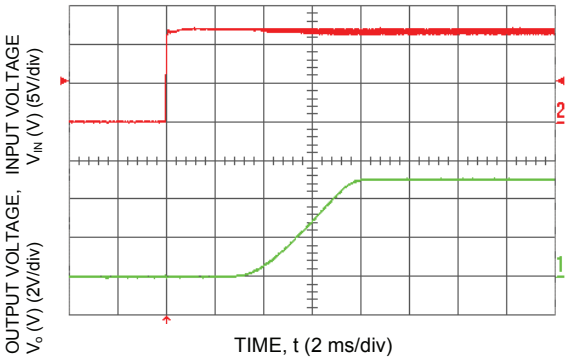


Figure 16. Typical Start-Up with application of Vin with low-ESR polymer capacitors at the output (7x150 µF) (Vin = 12Vdc, Vo = 5.0Vdc, Io = 16A, Co = 1050 µF).

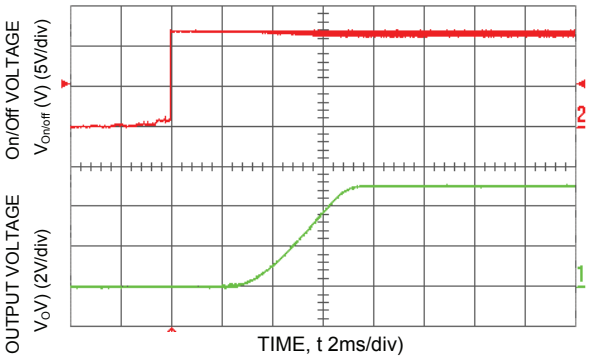


Figure 14. Typical Start-Up Using Remote On/Off (Vin = 12Vdc, Vo = 5.0Vdc, Io =16A).

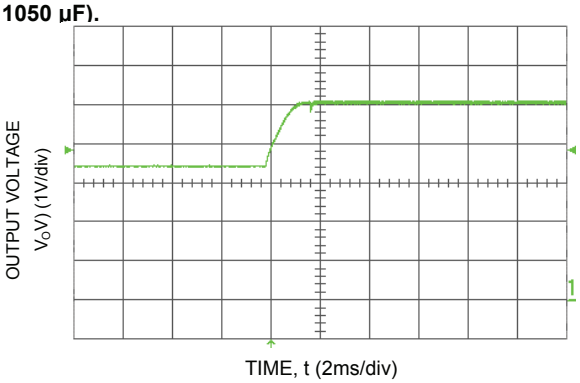


Figure 17. Typical Start-Up with Prebias (Vin = 12Vdc, Vo = 2.5Vdc, Io = 1A, Vbias =1.2 Vdc).

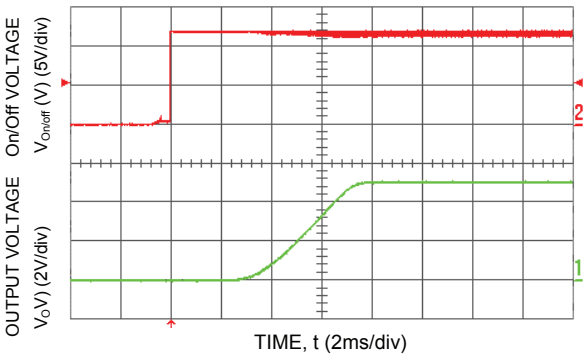


Figure 15. Typical Start-Up Using Remote On/Off with Low-ESR external capacitors (7x150uF Polymer) (Vin = 12Vdc, Vo = 5.0Vdc, Io = 16A, Co = 1050µF).

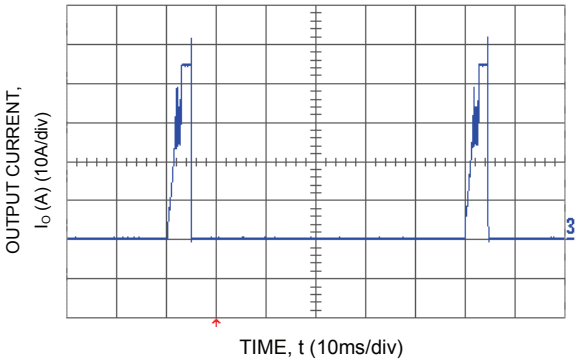


Figure 18. Output short circuit Current (Vin = 12Vdc, Vo = 0.75Vdc).

Characteristic Curves (continued)

The following figures provide thermal derating curves for the Austin SuperLynx™ II 12V SIP modules.

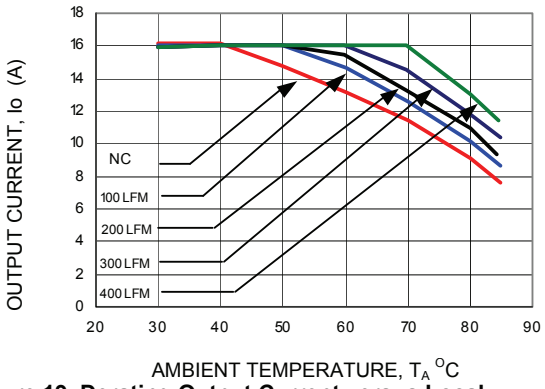


Figure 19. Derating Output Current versus Local Ambient Temperature and Airflow (Vin = 12Vdc, Vo=0.75Vdc).

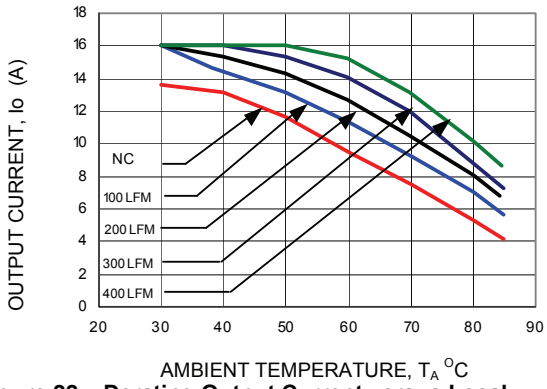


Figure 22. Derating Output Current versus Local Ambient Temperature and Airflow (Vin = 12dc, Vo=5.0 Vdc).

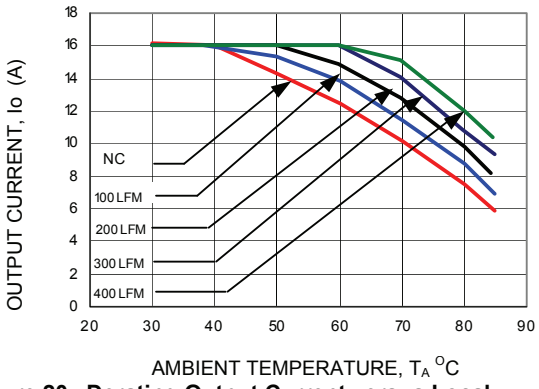


Figure 20. Derating Output Current versus Local Ambient Temperature and Airflow (Vin = 12Vdc, Vo=1.8 Vdc).

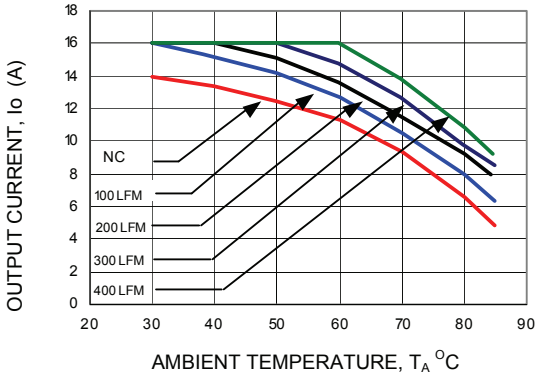
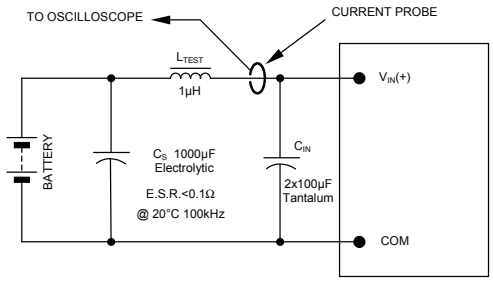


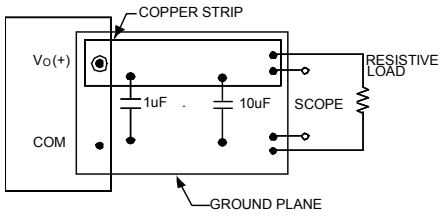
Figure 21. Derating Output Current versus Local Ambient Temperature and Airflow (Vin = 12Vdc, Vo=3.3 Vdc).

Test Configurations



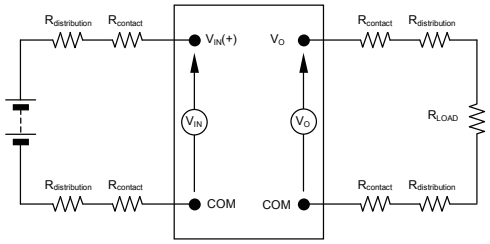
NOTE: Measure input reflected ripple current with a simulated source inductance (L_{TEST}) of 1µH. Capacitor C_S offsets possible battery impedance. Measure current as shown above.

Figure 23. Input Reflected Ripple Current Test Setup.



NOTE: All voltage measurements to be taken at the module terminals, as shown above. If sockets are used then Kelvin connections are required at the module terminals to avoid measurement errors due to socket contact resistance.

Figure 24. Output Ripple and Noise Test Setup.



NOTE: All voltage measurements to be taken at the module terminals, as shown above. If sockets are used then Kelvin connections are required at the module terminals to avoid measurement errors due to socket contact resistance.

Figure 25. Output Voltage and Efficiency Test Setup.

$$\text{Efficiency } \eta = \frac{V_O . I_O}{V_{IN} . I_{IN}} \times 100 \%$$

Design Considerations

Input Filtering

The Austin SuperLynx™ II 12V SIP module should be connected to a low-impedance source. A highly inductive source can affect the stability of the module. An input capacitance must be placed directly adjacent to the input pin of the module, to minimize input ripple voltage and ensure module stability.

In a typical application, 6x47 µF low-ESR tantalum capacitors (AVX part #: TPSE476M025R0100, 47µF 25V 100 mΩ ESR tantalum capacitor) will be sufficient to provide adequate ripple voltage at the input of the module. To further minimize ripple voltage at the input, very low ESR ceramic capacitors are recommended at the input of the module. Figure 26 shows input ripple voltage (mVp-p) for various outputs with 6x47 µF tantalum capacitors and with 6x22 µF ceramic capacitor (TDK part #: C4532X5R1C226M) at full load.

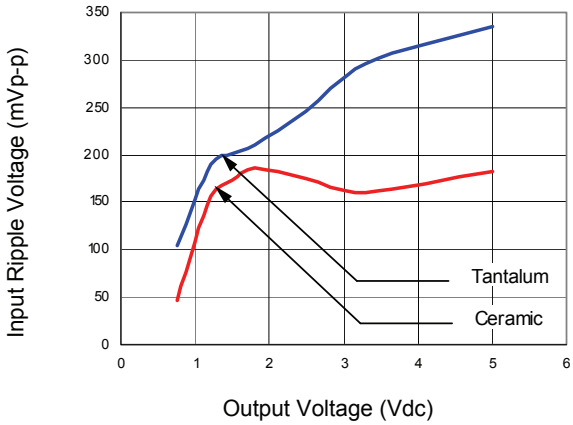


Figure 26. Input ripple voltage for various output with 6x47 µF tantalum capacitors and with 6x22 µF ceramic capacitors at the input (full load).

Design Considerations (continued)

Output Filtering

The Austin SuperLynx™ II 12V SIPmodule is designed for low output ripple voltage and will meet the maximum output ripple specification with 1 µF ceramic and 10 µF tantalum capacitors at the output of the module. However, additional output filtering may be required by the system designer for a number of reasons. First, there may be a need to further reduce the output ripple and noise of the module. Second, the dynamic response characteristics may need to be customized to a particular load step change.

To reduce the output ripple and improve the dynamic response to a step load change, additional capacitance at the output can be used. Low ESR polymer and ceramic capacitors are recommended to improve the dynamic response of the module. For stable operation of the module, limit the capacitance to less than the maximum output capacitance as specified in the electrical specification table.

Safety Considerations

For safety agency approval the power module must be installed in compliance with the spacing and separation requirements of the end-use safety agency standards, i.e., UL 60950-1, CSA C22.2 No. 60950-1-03, and VDE 0850:2001-12 (EN60950-1) Licensed.

For the converter output to be considered meeting the requirements of safety extra-low voltage (SELV), the input must meet SELV requirements. The power module has extra-low voltage (ELV) outputs when all inputs are ELV.

The input to these units is to be provided with a fast-acting fuse with a maximum rating of 6A in the positive input lead.

Feature Description

Remote On/Off

Austin SuperLynx™ II 12V SIP power modules feature an On/Off pin for remote On/Off operation. Two On/Off logic options are available in the Austin SuperLynx™ II series modules. Positive Logic On/Off signal, device code suffix "4", turns the module ON during a logic High on the On/Off pin and turns the module OFF during a logic Low. Negative logic On/Off signal, no device code suffix, turns the module OFF during logic High and turns the module ON during logic Low.

For positive logic modules, the circuit configuration for using the On/Off pin is shown in Figure 27. The On/Off pin is an open collector/drain logic input signal (Von/Off) that is referenced to ground. During a logic-high (On/Off pin is pulled high internal to the module) when the transistor Q1 is in the Off state, the power module is ON. Maximum allowable leakage current of the transistor when Von/off = $V_{IN,max}$ is 10μA. Applying a logic-low when the transistor Q1 is turned-On, the power module is OFF. During this state VOn/Off must be less than 0.3V. When not using positive logic On/off pin, leave the pin unconnected or tie to $V_{IN.}$

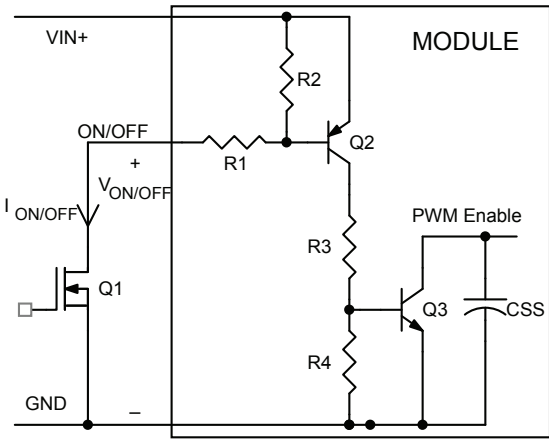


Figure 27. Circuit configuration for using positive logic On/OFF.

For negative logic On/Off devices, the circuit configuration is shown is Figure 28. The On/Off pin is pulled high with an external pull-up resistor (typical $R_{pull-up}$ = 68k, +/- 5%). When transistor Q1 is in the Off state, logic High is applied to the On/Off pin and the power module is Off. The minimum On/off voltage for logic High on the On/Off pin is 2.5 Vdc. To turn the module ON, logic Low is applied to the On/Off pin by turning ON Q1. When not using the negative logic On/Off, leave the pin unconnected or tie to GND.

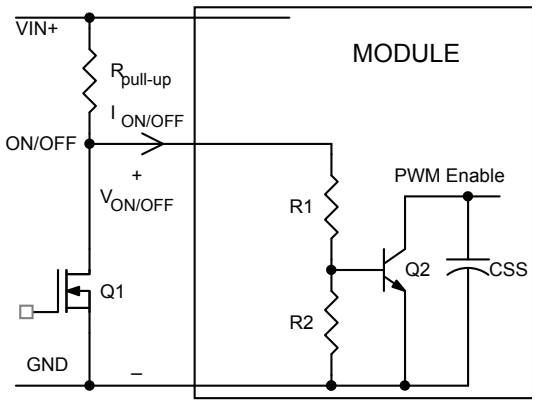


Figure 28. Circuit configuration for using negative logic On/OFF.

Overcurrent Protection

To provide protection in a fault (output overload) condition, the unit is equipped with internal current-limiting circuitry and can endure current limiting continuously. At the point of current-limit inception, the unit enters hiccup mode. The unit operates normally once the output current is brought back into its specified range. The typical average output current during hiccup is 3A.

Input Undervoltage Lockout

At input voltages below the input undervoltage lockout limit, module operation is disabled. The module will begin to operate at an input voltage above the undervoltage lockout turn-on threshold.

Overtemperature Protection

To provide protection in a fault condition, the unit is equipped with a thermal shutdown circuit. The unit will shutdown if the thermal reference point T_{ref}, exceeds 125°C (typical), but the thermal shutdown is not intended as a guarantee that the unit will survive temperatures beyond its rating. The module will automatically restarts after it cools down.

Feature Descriptions (continued)

Output Voltage Programming

The output voltage of the Austin SuperLynx™ II 12V can be programmed to any voltage from 0.75Vdc to 5.5Vdc by connecting a resistor (shown as *Rtrim* in Figure 29) between the Trim and GND pins of the module. Without an external resistor between the Trim and GND pins, the output of the module will be 0.7525Vdc. To calculate the value of the trim resistor, *Rtrim* for a desired output voltage, use the following equation:

$$Rtrim = \left[\frac{10500}{Vo - 0.7525} - 1000\right]\Omega$$

Rtrim is the external resistor in Ω

Vo is the desired output voltage

For example, to program the output voltage of the Austin SuperLynx™ II module to 1.8V, *Rtrim* is calculated as follows:

$$Rtrim = \left[\frac{10500}{1.8 - 0.75} - 1000\right]$$

$$Rtrim = 9.024k\Omega$$

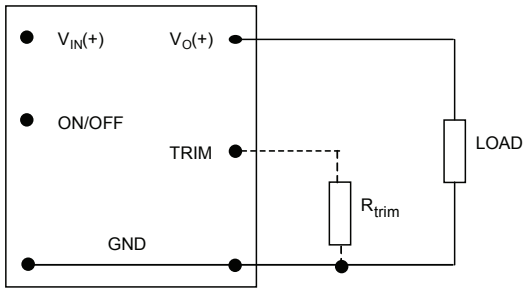


Figure 29. Circuit configuration to program output voltage using an external resistor

Table 1 provides *Rtrim* values for most common output voltages.

Table 1

$V_{O,\,set}$ (V)	*Rtrim* (KΩ)
0.7525	Open
1.2	22.46
1.5	13.05
1.8	9.024
2.5	5.009
3.3	3.122
5.0	1.472

By using 1% tolerance trim resistor, set point tolerance of ±2% is achieved as specified in the electrical specification. The POL Programming Tool, available at www.lineagepower.com under the Design Tools section, helps determine the required external trim resistor needed for a specific output voltage.

The amount of power delivered by the module is defined as the voltage at the output terminals multiplied by the output current. When using the trim feature, the output voltage of the module can be increased, which at the same output current would increase the power output of the module. Care should be taken to ensure that the maximum output power of the module remains at or below the maximum rated power ($P_{max} = V_{o,set} \times I_{o,max}$).

Voltage Margining

Output voltage margining can be implemented in the Austin SuperLynx™ II modules by connecting a resistor, $R_{margin\text{-}up}$, from the Trim pin to the ground pin for margining-up the output voltage and by connecting a resistor, $R_{margin\text{-}down}$, from the Trim pin to the Output pin for margining-down. Figure 30 shows the circuit configuration for output voltage margining. The POL Programming Tool, available at www.lineagepower.com under the Design Tools section, also calculates the values of $R_{margin\text{-}up}$ and $R_{margin\text{-}down}$ for a specific output voltage and % margin. Please consult your local Lineage Power technical representative for additional details.

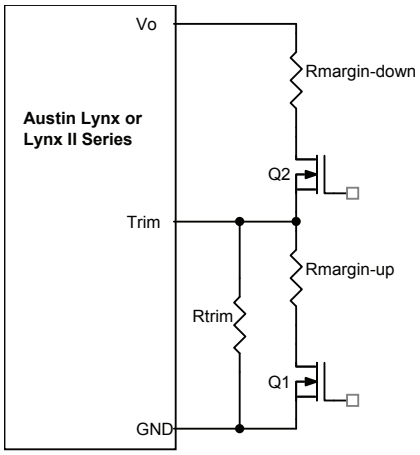


Figure 30. Circuit Configuration for margining Output voltage.

Feature Descriptions (continued)

Voltage Sequencing

Austin SuperLynx™ II 12V series of modules include a sequencing feature, EZ-SEQUENCE™ that enables users to implement various types of output voltage sequencing in their applications. This is accomplished via an additional sequencing pin. When not using the sequencing feature, either tie the SEQ pin to V_{IN} or leave it unconnected.

When an analog voltage is applied to the SEQ pin, the output voltage tracks this voltage until the output reaches the set-point voltage. The SEQ voltage must be set higher than the set-point voltage of the module. The output voltage follows the voltage on the SEQ pin on a one-to-one volt basis. By connecting multiple modules together, customers can get multiple modules to track their output voltages to the voltage applied on the SEQ pin.

For proper voltage sequencing, first, input voltage is applied to the module. The On/Off pin of the module is left unconnected (or tied to GND for negative logic modules or tied to V_{IN} for positive logic modules) so that the module is ON by default. After applying input voltage to the module, a minimum of 10msec delay is required before applying voltage on the SEQ pin. During this time, potential of 50mV (± 10 mV) is maintained on the SEQ pin. After 10msec delay, an analog voltage is applied to the SEQ pin and the output voltage of the module will track this voltage on a one-to-one volt bases until output reaches the set-point voltage. To initiate simultaneous shutdown of the modules, the SEQ pin voltage is lowered in a controlled manner. Output voltage of the modules tracks the voltages below their set-point voltages on a one-to-one basis. A valid input voltage must be maintained until the tracking and output voltages reach ground potential to ensure a controlled shutdown of the modules.

When using the EZ-SEQUENCE™ feature to control start-up of the module, pre-bias immunity feature during start-up is disabled. The pre-bias immunity feature of the module relies on the module being in the diode-mode during start-up. When using the EZ-SEQUENCE™ feature, modules goes through an internal set-up time of 10msec, and will be in synchronous rectification mode when voltage at the SEQ pin is applied. This will result in sinking current in the module if pre-bias voltage is present at the output of the module. When pre-bias immunity during start-up is required, the EZ-SEQUENCE™ feature must be disabled. For additional guidelines on using EZ-SEQUENCE™ feature of Austin SuperLynx™ II 12V, contact Lineage Power technical representative for preliminary application note on output voltage sequencing using Austin Lynx II series.

Remote Sense

The Austin SuperLynx™ II 12V SIP power modules have a Remote Sense feature to minimize the effects of distribution losses by regulating the voltage at the Remote Sense pin (See Figure 31). The voltage between the Sense pin and Vo pin must not exceed 0.5V.

The amount of power delivered by the module is defined as the output voltage multiplied by the output current (Vo x Io). When using Remote Sense, the output voltage of the module can increase, which if the same output is maintained, increases the power output by the module. Make sure that the maximum output power of the module remains at or below the maximum rated power. When the Remote Sense feature is not being used, connect the Remote Sense pin to output pin of the module.

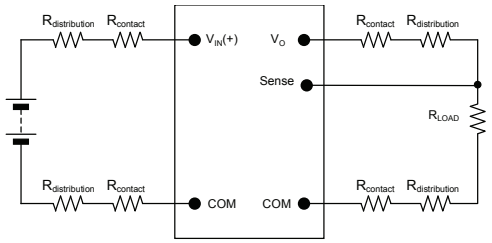


Figure 31. Remote sense circuit configuration.

Thermal Considerations

Power modules operate in a variety of thermal environments; however, sufficient cooling should be provided to help ensure reliable operation.

Considerations include ambient temperature, airflow, module power dissipation, and the need for increased reliability. A reduction in the operating temperature of the module will result in an increase in reliability. The thermal data presented here is based on physical measurements taken in a wind tunnel. The test set-up is shown in Figure 33. Note that the airflow is parallel to the long axis of the module as shown in figure 32. The derating data applies to airflow in either direction of the module's long axis.

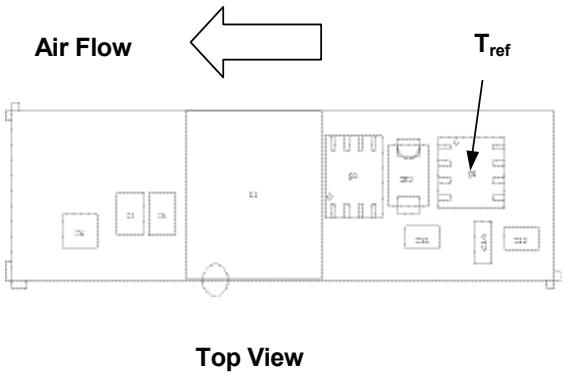


Figure 32. T_{ref} Temperature measurement location.

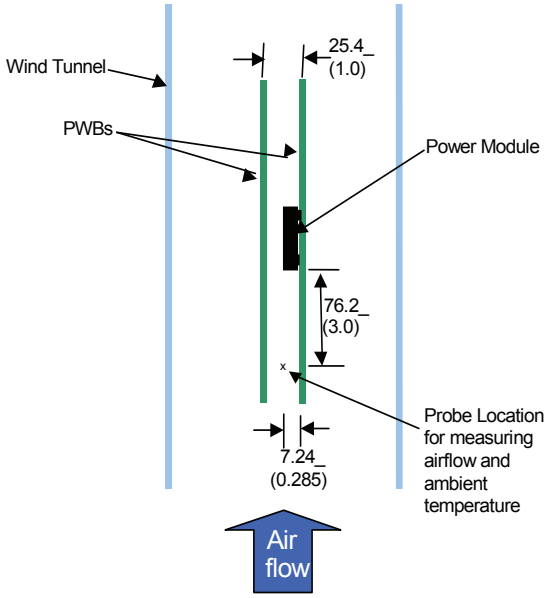


Figure 33. Thermal Test Set-up.

The thermal reference point, $T_{ref\,1}$ used in the specifications of thermal derating curves is shown in Figure 32. For reliable operation this temperature should not exceed 125°C.

The output power of the module should not exceed the rated power of the module (Vo,set x Io,max).

Please refer to the Application Note "Thermal Characterization Process For Open-Frame Board-Mounted Power Modules" for a detailed discussion of thermal aspects including maximum device temperatures.

Heat Transfer via Convection

Increased airflow over the module enhances the heat transfer via convection. Thermal derating curves showing the maximum output current that can be delivered by various module versus local ambient temperature (T_A) for natural convection and up to 1m/s (200 ft./min) are shown in the Characteristics Curves section.

Post solder Cleaning and Drying Considerations

Post solder cleaning is usually the final circuit-board assembly process prior to electrical board testing. The result of inadequate cleaning and drying can affect both the reliability of a power module and the testability of the finished circuit-board assembly. For guidance on appropriate soldering, cleaning and drying procedures, refer to *Board Mounted Power Modules: Soldering and Cleaning* Application Note.

Through-Hole Lead-Free Soldering Information

The RoHS-compliant through-hole products use the SAC (Sn/Ag/Cu) Pb-free solder and RoHS-compliant components. They are designed to be processed through single or dual wave soldering machines. The pins have an RoHS-compliant finish that is compatible with both Pb and Pb-free wave soldering processes. A maximum preheat rate of 3°C/s is suggested. The wave preheat process should be such that the temperature of the power module board is kept below 210°C. For Pb solder, the recommended pot temperature is 260°C, while the Pb-free solder pot is 270°C max. Not all RoHS-compliant through-hole products can be processed with paste-through-hole Pb or Pb-free reflow process. If additional information is needed, please consult with your Lineage Power technical representative for more details.

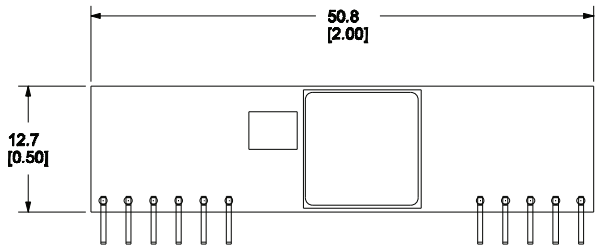
Mechanical Outline

Dimensions are in millimeters and (inches).

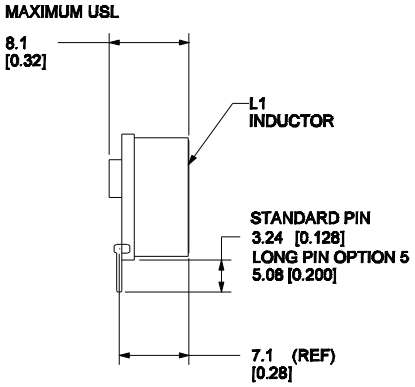
Tolerances: x.x mm ± 0.5 mm (x.xx in. ± 0.02 in.) [unless otherwise indicated]

x.xx mm ± 0.25 mm (x.xxx in ± 0.010 in.)

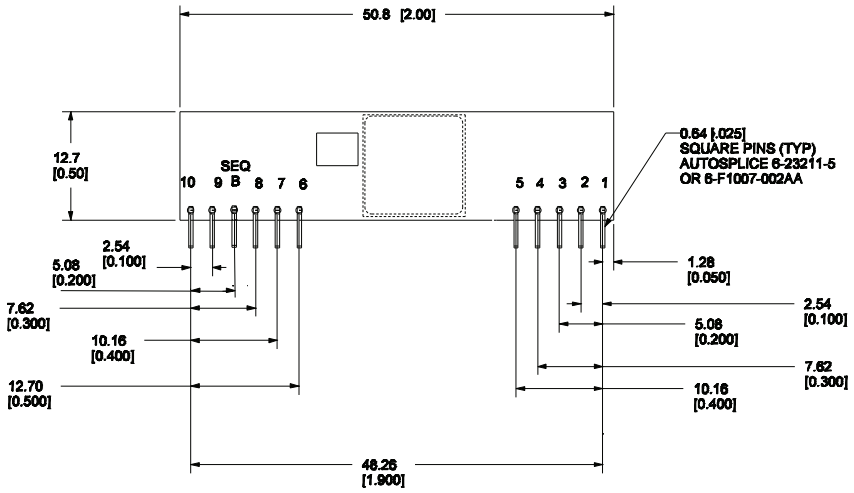
Top View



Side View



Bottom View



PIN	FUNCTION
1	Vo
2	Vo
3	Sense+
4	Vo
5	GND
6	GND
7	VIN
8	VIN
B	SEQ
9	Trim
10	On/Off

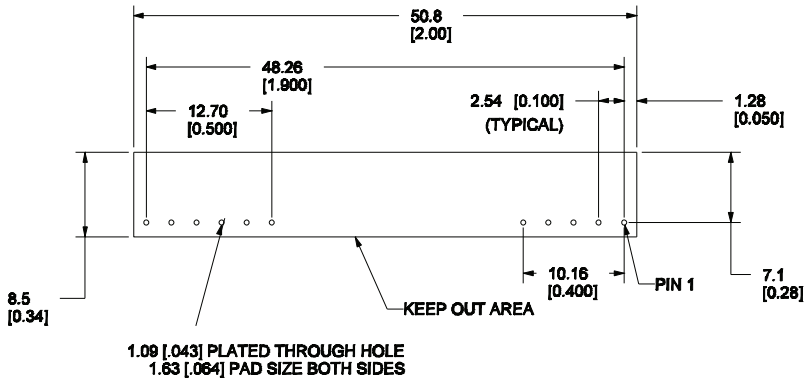
Recommended Pad Layout

Dimensions are in millimeters and (inches).

Tolerances: x.x mm ± 0.5 mm (x.xx in. ± 0.02 in.) [unless otherwise indicated]

x.xx mm ± 0.25 mm (x.xxx in ± 0.010 in.)

PIN	FUNCTION
1	Vo
2	Vo
3	Sense+
4	Vo
5	GND
6	GND
7	VIN
8	VIN
B	SEQ
9	Trim
10	On/Off



Through- Hole Pad Layout – Back view

Ordering Information

Please contact your Lineage Power Sales Representative for pricing, availability and optional features.

Table 2. Device Codes

Device Code	Input Voltage	Output Voltage	Output Current	Efficiency 3.3V@ 16A	Connector Type	Comcodes
ATA016A0X3	8.3 – 14Vdc	0.75 – 5.5Vdc	16 A	92.0%	SIP	108989091
ATA016A0X3Z	8.3 – 14Vdc	0.75 – 5.5Vdc	16 A	92.0%	SIP	CC109104691
ATA016A0X43	8.3 – 14Vdc	0.75 – 5.5Vdc	16 A	92.0%	SIP	108989100
ATA016A0X43Z	8.3 – 14Vdc	0.75 – 5.5Vdc	16 A	92.0%	SIP	CC109104700

-Z refers to RoHS-compliant versions.

Table 3. Device Option

Option*	Suffix**
Long Pins 5.08 mm ± 0.25mm (0.200 in. ± 0.010 in.)	5

* Contact Lineage Power Sales Representative for availability of these options, samples, minimum order quantity and lead times

** When adding multiple options to the product code, add suffix numbers in the descending order

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